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### **Optical Sensors**

Application Note

# **Designing the VCNL36829UM Into an Application**

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### ABSTRACT

This application note provides an introduction to the functionality of the VCNL36829UM sensor, application circuits, and mechanical design considerations.

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## **Designing the VCNL36829UM Into an Application**

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## **Designing the VCNL36829UM Into an Application**

### **1. INTRODUCTION**

The VCNL36829UM is a fully integrated proximity sensor. It combines a vertical-cavity surface-emitting laser (VCSEL), photodiode, and application-specific integrated circuit (ASIC) within a single package. The VCNL36829UM has been developed for proximity detection applications that require a dual slave address, low power consumption, small package size, small window size, and short range operation up to 5 cm. In addition, given the rated supply voltage of 1.8 V to reduce power consumption, the sensor is intended for battery-powered applications such as:

- Presence detection in
  - Smartphones and true wireless stereo (TWS) earbuds
  - VR / AR headsets and smart glasses
  - Smartwatches
- Touchless button / dispensing



Fig. 1 - The VCNL36829UM

KEY BENEFITS OF USING VCNL36829UM					
BENEFITS	DESCRIPTION				
Small package for	Very small 1.6 x 1.0 x 0.35 (L x W x H in mm) package				
tight space requirements	A small package allows a design with a small window size				
	A 1.8 V rated power supply and I <sup>2</sup> C bus reduce the power consumption				
Low power consumption for battery-powered devices	The low idle current of 5 $\mu$ A allows a lower average current consumption				
	The use of VCSEL instead of IRLED allows a lower driving current with comparable performance				
Superior provinity detection	The proximity sensor can detect up to 5 cm distance				
Superior proximity detection	The proximity sensor supports sunlight cancellation up to 200 klux of sunlight				
Smart dual slave address	The smart dual slave address allows the connection of up to two proximity sensors with a microcontroller, without needing an I <sup>2</sup> C multiplexer; the slave address can be changed by swapping the SCL and SDA pins				

This application note describes the functionality, application circuits, register settings, and mechanical design considerations for the sensor.



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### 2. PIN DESCRIPTION AND BLOCK DIAGRAM

2.1 Pin Description



Notes

(1) ASIC - Application-specific integrated circuit

<sup>(2)</sup> PD - Photodiode

(3) VCSEL - Vertical-cavity surface-emitting laser

Fig. 2 - Pin Assignment (top view) of the VCNL36829UM

TABLE 1 - PIN DESCRIPTION								
PIN NUMBER	PIN NAME	TYPE	DESCRIPTION					
1	INT	O (open drain)	Interrupt					
2	V <sub>DD</sub>	I	Supply voltage					
3	GND	I	Ground					
4	VCSELA	I	VCSEL anode					
5	SCL (1)	I / O (open drain)	I <sup>2</sup> C serial clock					
6	SDA <sup>(1)</sup>	I / O (open drain)	l <sup>2</sup> C serial data					

Note

<sup>(1)</sup> Pin 5 (SCL) and pin 6 (SDA) can be swapped to change the slave address from 0x60 to 0x51 (7-bit slave address)



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#### 2.2 Block Diagram



Fig. 3 - Block Diagram of the VCNL36829UM

TABLE 2 - BLOCK DIAGRAM DESCRIPTION						
COMPONENT	DESCRIPTION					
Command registers	The command registers are the memory storage for writing and reading I <sup>2</sup> C commands					
I <sup>2</sup> C interface	The I <sup>2</sup> C interface is a communication interface with active low open drain circuitry					
Interrupt control	Interrupt control is a circuit block with active low open drain output					
PD	PD is a photodiode that converts the reflected infrared signal from the object into photocurrent; it is then fed to the 16-bit ADC					
16-bit ADC	The 16-bit analog to digital converter converts the analog signal to the digital signal <sup>(1)</sup> ; the input signal is then amplified					
Logic and timing controller	The logic and timing controller controls the timing for the proximity measurement					
VCSEL driver	The VCSEL driver is a circuitry block that limits the driving current based on the selected setting					
Oscillator	The oscillator generates the clock signal to synchronize all of the device functionalities					

#### Note

(1) The actual bit resolution can vary between 14 bits and 16 bits, depending on the register settings PS\_IT. If PS\_IT = 12.5 μs to 175 μs, the full code is 14 bit /16 383 counts; if PS\_IT = 200 μs to 375 μs, the full code is 16 bit / 65 535 counts

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### 3. APPLICATION CIRCUIT

#### 3.1 Slave Address Selection

The VCNL36829UM supports a smart dual slave address in which the designer can change the slave address by swapping the SCL and SDA pins, as shown in Table 3.

TABLE 3 - SLAVE ADDRESS TABLE								
PIN 5	PIN 6	7-BIT SLAVE ADDRESS	8-BIT SLAVE ADDRESS (WRITE)	8-BIT SLAVE ADDRESS (READ)				
SCL	SDA	0x60	0xC0	0xC1				
SDA	SCL	0x51	0xA2	0xA3				

A smart dual slave address provides the flexibility for the designer to connect two devices from two different slave addresses on the same I<sup>2</sup>C bus. The two slave address options allow designers to select a different slave address if one is used by the other slave devices on the same I<sup>2</sup>C bus in a single device application.

#### 3.2 Application Circuit With a Single Device

Fig. 4 and Fig. 5 show application circuit examples with a single device. As described in Table 3, when pins 5 and 6 are connected to the clock and data signal from the microcontroller, as shown in Fig. 4, they will then be configured as an SCL pin and SDA pin, respectively. The 7-bit slave address option of 0x60 will be automatically selected.



Fig. 4 - Application Circuit Example for a Single VCNL36829UM - Slave Address 0x60

On the other hand, when pins 5 and 6 are connected to the data and clock signal from the microcontroller, as shown in Fig. 5, they will then be configured as an SDA pin and SCL pin, respectively. The 7-bit slave address option of 0x51 will be automatically selected.



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Fig. 5 - Application Circuit Example for a Single VCNL36829UM - Slave Address 0x51

Table 4 shows the required values and the explanation for the individual application circuit parameters.

TABLE 4 - APPLICATION CIRCUIT PARAMETERS						
CIRCUIT PARAMETER	VALUE	DESCRIPTION				
V <sub>DD</sub>	1.65 V to 2.00 V	A stable power supply is recommended, such as a low dropout or switching regulator; the power supply isolation can be further improved with a decoupling capacitor $C_1$				
V <sub>VCSEL</sub>	2.80 V to 3.60 V	A stable power supply such as a low dropout or switching regulator that can supply an adequate amount of power (max. VCSEL pulse driving current of 18 mA) is recommended; the power supply isolation can be further improved with a decoupling capacitor $C_2$ ; the minimum voltage depends on the selected driving current of the VCSEL; please refer to Table 5 for reference				
V <sub>PULL UP</sub>	1.2 V to 3.6 V	A stable power supply such as a low dropout or switching regulator is recommended; a voltage level shifter is required if the I <sup>2</sup> C bus voltage from the microcontroller is higher than 3.6 V				
C <sub>1</sub> to C <sub>4</sub>	100 nF to 1 µF	Decoupling capacitors are recommended to reduce the noise in the supply voltage				
R <sub>1</sub> to R <sub>2</sub>	2.2 k $\Omega$ to 4.7 k $\Omega$	Pull-up resistors within the range of 2.2 k $\Omega$ to 4.7 k $\Omega$ are recommended; any increase in bus capacitance or resistance will increase the logic-high transition time				
R <sub>3</sub>	4.7 k $\Omega$ to 22 k $\Omega$	A pull-up resistor within the range of 4.7 $k\Omega$ to 22 $k\Omega$ is recommended				

#### VCSEL Driver

Fig. 6 shows the VCSEL driver circuit in the VCNL36829UM.



The typical supply voltage for the VCSEL is 3.3 V. The designer could connect pin 4 VCSEL<sub>A</sub> with a lower supply voltage  $V_{VCSEL}$ . However,  $V_{VCSEL}$  should at least match the minimum required supply voltage for the VCSEL  $V_{VCSEL.min}$ .

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V<sub>VCSEL,min.</sub> depends on:

- $\bullet$  The selected driving current  $I_{\text{F}}$  of the VCSEL and its corresponding forward voltage  $V_{\text{F}}$
- $\bullet$  The forward voltage tolerance of the VCSEL,  $V_{F,\ tolerance},$  which is 0.25 V
- $\bullet$  The minimum voltage required by the driver circuit,  $V_{\text{Driver},\text{min.}},$  which is 0.6 V

Therefore,

$$V_{VCSEL, min.} = V_F(I_F) + V_{F, tolerance} + V_{Driver, min.} = V_F(I_F) + 0.25 V + 0.6 V$$

Fig. 7 and Table 5 show the computation of  $V_{VCSEL,min.}$ . The designer could use 3.3 V to provide sufficient voltage headroom. The maximum allowable supply voltage of the VCSEL  $V_{VCSEL,max.}$  is 3.6 V.



Fig. 7 - Forward Current vs. VCSEL Supply Voltage

TABLE 5 - V <sub>VCSEL, MIN.</sub>									
I <sub>VCSEL</sub> (I <sub>F</sub> )	8 mA	8 mA 10 mA 12 mA 14 mA 16 mA 18 mA							
V <sub>F</sub>	1.95 V	2.05 V	2.15 V	2.23 V	2.31 V	2.37 V			
V <sub>F, tolerance</sub>			0.2	5 V					
V <sub>Driver, min.</sub>			0.6	5 V					
V <sub>VCSEL, min.</sub>	2.80 V 2.90 V 3.00 V 3.08 V 3.16 V 3.22 V								
V <sub>VCSEL, max.</sub>	3.6 V								

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#### 3.3 Application Circuit With a Smart Dual Slave Address

Fig. 8 shows an application circuit example with a smart dual slave address. By swapping the SCL and SDA pins of the second device, as shown in Table 3, the designer can change the 7 bit slave address of the VCNL36829UM. This provides the flexibility for the designer to connect two devices from two different slave addresses on the same I<sup>2</sup>C bus.



Fig. 8 - Application Circuit Example for Two VCNL36829UMs - Smart Dual Slave Address



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### 4. REGISTER DESCRIPTION

TABLE 6 - REGISTER DESCRIPTION OVERVIEW									
REGISTER GROUP	BIT NAME	REGISTER NAME	COMMAND CODE	DESCRIPTION	ACCESS				
	PS_CAL			Enable / disable the internal calibration					
Basic	PS_TRIG	PS_CONF1_L		Set the active force mode trigger; this bit will be reset to 0 after the measurement cycle					
initialization	PS_MODE		0x00	Set the measurement mode of the sensor					
	PS_ON			Switch the sensor on / off					
	PS_SD	PS_CONF1_H		PS shutdown setting					
Emitter	PS _IT	PS_CONF2_H	0x01	Set the integration time for one measurement					
(VCSEL)	PS_CURRENT_EN			VCSEL driver enable setting					
settings	PS_CURRENT	PS_CONF3_L		Set the VCSEL driving current					
	PS_HG			Set the gain of the ADC					
	PS_GAIN		0x02	Set the gain of the ADC					
Detector	PD3_EN	PS_CONF3_H		PD3 enable setting					
Settings	PD2_EN			PD2 enable setting					
	PD1_EN			PD1 enable setting					
Wait time setting	PS_WAIT	PS_CONF2_L	0x01	PS wait time after PS detection					
-	Reserved	PS_CONF4_L		Reserved	Write				
Sunlight PS_SC_LEVEL cancellation PS_SC		PS_CONF4_H	0x03	Sunlight cancellation level setting	read				
				Enable / disable the sunlight cancellation					
Offset count	et count PS_CANC_L PS_CANC_L 000 O		Offset count cancellation value setting (low byte)						
cancellation	tion PS_CANC_H PS_CANC_H 0X06		0x06	Offset count cancellation value setting (high byte)					
- Reserved		PS_CONF5_L PS_CONF5_H	0x08	Reserved (must be set to 0xA0)					
				Reserved					
	PS_INT				Set the interrupt mode setting				
	PS_PERS			Set the amount of consecutive threshold crossing events necessary to trigger interrupt					
	PS_SMART_PERS	PS_CONF1_H	0x00	Enable / disable the smart persistence setting when the interrupt event is triggered					
	PS_SP_INT			Enable / disable the sunlight protection mode interrupt setting					
	PS_START_INT			After PS_SD disable, PS 1 <sup>st</sup> detection Interrupt enable setting					
Interrupt	PS_THDL_L	PS_THDL_L	0×04	Low threshold interrupt value setting (low byte)					
	PS_THDL_H	PS_THDL_H	0,04	Low threshold interrupt value setting (high byte)					
	PS_THDH_L	PS_THDH_L	0×05	High threshold interrupt value setting (low byte)					
	PS_THDH_H	PS_THDH_H	0,03	High threshold interrupt value setting (high byte)					
	PS_START_FALG			PS finish 1st detection after SD disable					
	PS_SPFLAG		0.450	Read the sunlight protection mode interrupt event flag					
	PS_IF_CLOSE	INT_FLAG	01 9	Read the high threshold crossing interrupt event flag					
	PS_IF_AWAY			Read the low threshold crossing interrupt event flag					
	PS_DATA_L		0.459	Proximity output data (low byte)	Read				
	PS_DATA_H	FS_DATA	UXFO	Proximity output data (high byte)	only				
Readout	VCNL36829UM_			Device ID <sup>(1)</sup> ; slave address: 0x60, ID = 0x29;	]				
registers	ID_L	VCNL36829UM	0xFA	slave address: 0x51, ID = 0x29					
	VCNL36829UM_ ID_H	_ID		Device ID <sup>(1)</sup> ; slave address: $0x60$ , ID = $0x00$ ; slave address: $0x51$ , ID = $0x10$					

Notes

(1) The default ID depends on the connection of the SCL and SDA pins on the VCNL36829UM with the SCL and SDA pins on the host MCU; if pins 5 and 6 on the VCNL36829UM are connected to the SCL and SDA pins on the host, the default value will be 0x0029; on the other hand, if pins 5 and 6 on the VCNL36829UM are connected to the SDA and SCL pins on the host, the default value will be 0x1029

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APPLICATION NOT

<sup>•</sup> All of the reserved registers are used for internal testing; these values must be kept constant



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#### 4.1 Basic Initialization

The sensor can be initialized with the PS\_CAL, PS\_ON, PS\_MODE, and PS\_TRIG bits, which are found in three different registers: PS\_CONF1\_L.

TABLE 7 - BASIC INITIALIZATION									
REGISTER GROUP	BIT NAME	REGISTER NAME	COMMAND CODE	FUNCTION		VALUE	DESCRIPTION		
	PS CAL			Enable / disable the internal calibration	7	0x0 (0b0)	Disable (default)		
	F3_UAL				1	0x1 (0b1)	Enable		
	Reserved		0x00	Reserved	6	0x0 (0b0)	Should be kept default		
PS.				Set the active force mode trigger; this bit will be reset to 0 after the measurement cycle	_	0x0 (0b0)	Off (default)		
	PS_TRIG				5	0x1 (0b1)	Trigger		
Basic		PS_CONF1_L		Set the measurement mode of the sensor	4	0x0 (0b0)	Auto mode (default)		
initialization	F3_IVIODE	-				0x1 (0b1)	Active force mode		
	Reserved			Reserved	3:1	0x0 (0b000)	Should be kept default		
	PS_ON			Switch the sensor on / off	0	0x0 (0b0)	Turn off the sensor (shutdown) (default)		
						0x1 (0b1)	Turn on the sensor		
			Ī	PS shutdown setting	8	0x0 (0b0)	Disable		
	P5_5D			PS shutdown setting	0	0x1 (0b1)	Enable (default)		

**PS\_ON** - The sensor can be turned on by setting this bit to 1 and turned off by setting it to 0. The sensor will mostly be in the idle state and only be in the active state during the measurement phase.

**PS\_MODE** - Set this bit to 0 to activate the auto mode and 1 to activate the active force mode. Auto mode means the measurement will be triggered continuously with a rate defined by the measurement period. This measurement period depends on the PS\_IT and PS\_WAIT bits' setting. Active force mode means the measurement will have to be triggered manually by setting the PS\_TRIG bit to 1.

**PS\_TRIG** - This bit must be set to 1 when using active force mode to trigger a measurement. Otherwise, the bit must be set to 0. **PS\_CAL** - Set the sensor to factory calibration values, when set to "1", ensuring original accuracy and stability. This bit will automatically return to "0".

Fig. 9 and Fig. 10 show the basic initialization of the two available modes. The basic initialization steps are usually useful when first testing with the proximity sensor. The proximity measurement will be based on the default values. In practice, more sensor parameters to change the strength of the VCSEL infrared signal, the gain of the ADC, the measurement period, and the interrupt should be set before starting the measurement.

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Fig. 9 - Basic Initialization Example Steps for Auto Mode

Fig. 10 - Basic Initialization Example Steps for Active Force Mode

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### 4.2 Emitter Settings (VCSEL)

TABLE 8 - EMITTER SETTINGS (VCSEL)								
REGISTER GROUP	BIT NAME	REGISTER NAME	COMMAND CODE	FUNCTION	BIT	VALUE	DESCRIPTION	
	Reserved			Reserved	15 : 12	0x0 (0b0000)	Should be kept default	
	PQ IT	PS_CONF2_H	0x01	Set the integration time	11 · 8	0x0 (0b0000)	12.5 µs (default)	
	F3_II			for one measurement	11.0	1 to 15	25 µs x PS_IT	
	Reserved		0x02	Reserved	7:4	0x0 (0b0000)	Should be kept default	
	PS_CURRENT_EN	PS_CONF3_L		VCSEL driver enable setting	3	0x0 (0b0)	Disable (default)	
						0x1 (0b1)	Enable	
Emitter					2:0	0x0 (0b000)	Reserved (default)	
settings						0x1 (0b001)	Reserved	
<b>J</b>						0x2 (0b010)	8 mA	
				Set the VCSEL		0x3 (0b011)	10 mA	
	F3_CONNENT			driving current		0x4 (0b100)	12 mA	
						0x5 (0b101)	14 mA	
						0x6 (0b110)	16 mA	
						0x7 (0b111)	18 mA	

Fig. 11 and Fig. 12 show the infrared signal of the VCSEL is a pulse with a width of T. This can be set via the parameter PS\_IT. The length of T can be 12.5  $\mu$ s to 375  $\mu$ s. The magnitude of the pulse can be controlled via the VCSEL driver current parameter PS\_CURRENT.

#### **PS\_CURRENT** - The magnitude of the driving current of the VCSEL pulse.



Fig. 11 - The Behavior of the Pulse Magnitude With Different PS\_CURRENT

### **PS\_IT** - The length of the VCSEL pulse width.



Fig. 12 - The Behavior of the Pulse Length With Different PS\_IT

Fig. 13 and Fig. 14 show the behavior of the proximity counts as the setting values increase. It can be observed that the O proximity counts increase significantly as the setting values are raised. If a longer detection distance is required, the designer Z could increase PS\_IT and PS\_CURRENT. The maximum proximity value is influenced by PS\_IT only, which will be explained further in Table 10.



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4.3 Detector Settings

TABLE 9	TABLE 9 - DETECTOR SETTINGS									
REGISTER GROUP	BIT NAME	REGISTER NAME	COMMAND CODE	FUNCTION	BIT	VALUE	DESCRIPTION			
	Reserved		0.00	Reserved	15	0x0 (0b0)	Should be kept default			
				Sat the gain of the ADC	1/	0x0 (0b0)	x1 gain (default)			
	P5_HG			Set the gain of the ADC	14	0x1 (0b1)	x2 gain			
				Set the gain of the ADC		0x0 (0b00)	x4 gain (default)			
	DS CAIN				12 . 10	0x1 (0b01)	x2 gain			
	10_0/014				13.12	0x2 (0b10)	x1 gain			
Detector						0x3 (0b11)	Reserved			
settings	Reserved	F3_00NF3_F	0,02	Reserved	11	0x0 (0b0)	Should be kept default			
				PD2 onable sotting	10	0x0 (0b0)	Disable (default)			
	FD3_EN			PD3 enable setting	10	0x1 (0b1)	Enable			
				PD2 onable sotting	0	0x0 (0b0)	Disable (default)			
	FDZ_LN			FD2 enable setting	9	0x1 (0b1)	Enable			
				PD1 onable setting	0	0x0 (0b0)	Disable (default)			
	FDI_EN			FDT enable setting	0	0x1 (0b1)	Enable			

The detector of the proximity sensor consists of the photodiode and the ADC. The detector-related parameters PS\_HG, PS\_GAIN, PD3\_EN, PD2\_EN, and PD1\_EN in register PS\_CONF3\_H double or halve the magnitude of proximity counts given the same reflected infrared signal magnitude from the object arrived at the detector. These parameters will help increase the number of counts when a higher dynamic range is needed. Fig. 15 depicts the behavior of the proximity counts when changing the ADC parameters.



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Fig. 15 - The Behavior of the Proximity Counts When Changing the ADC Parameters

**PD1\_EN / PD2\_EN / PD3\_EN** - This is related to the PD area configuration, where PD3 is twice the size of PD1 and PD2. Please refer to Fig. 18. PD settings can be independently enabled or disabled. PD1\_EN and PD2\_EN have similar counts, while PD3\_EN is double. At least one PD must be enabled to generate counts, and they can be used individually or combined for cumulative counting.

**PS\_GAIN** - This bit sets the gain. When the bit is set to 0, the counts are quadrupled. When combined with PS\_HG, if PS\_GAIN is set to 0 and PS\_HG is set to 1, the counts can be increased to eight times.

**PS\_HG** - This bit sets the sensitivity in the ADC. The counts can be doubled when the bit is set to 1. When combined with PS\_GAIN, the count can be increased to eight times when PS\_GAIN is set to 0 and PS\_HG is set to 1. The actual maximum resolution depends on PS\_IT, as shown in Table 10.

TABLE 10 - MAXIMUM BIT RESOLUTION AND DIGITAL OUTPUT COUNTS							
BIT	NAME	PS_IT = 12.5 μs to 175 μs	PS_IT = 200 µs to 375 µs				
	PS_GAIN = 00 (x4 gain)						
$PS_HG = 0$	PS_GAIN = 01 (x2 gain)						
	PS_GAIN = 10 (x1 gain)	14 hit (16 202 counts					
	PS_GAIN = 00 (x4 gain)	14 bit / 16 383 counts	16 DIT / 65 535 COUNTS				
PS_HG = 1	PS_GAIN = 01 (x2 gain)						
	PS_GAIN = 10 (x1 gain)						

Note

• Of the three PD\_EN settings, at least one must be set to 1, all cannot be 0; it is recommended to enable all (set to 1) to ensure adequate sensitivity

In many applications where the detection range is less than 5 cm, such as presence detection in TWS, AR / VR headsets, and touchless switches, a 14-bit resolution with maximum counts of 16 383 is usually enough to detect two distinct events of "close" and "away".

Fig. 16 and Fig. 17 show the behavior of the parameters PS\_GAIN and PS\_HG on the proximity counts. The proximity counts will double with an increase in the gain or sensitivity of the ADC. The combination of double gain and high sensitivity allows the proximity counts to be quadrupled.



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on the Proximity Counts



Fig. 17 - The Behavior of the Parameter PS\_HG on the Proximity Counts





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#### 4.4 Wait Time Settings

TABLE 11 - WAIT TIME SETTINGS								
REGISTER GROUP	BIT NAME	REGISTER NAME	COMMAND CODE	FUNCTION	BIT	VALUE	DESCRIPTION	
Wait time settings	PS_WAIT	PS_CONF2_L	0x01	PS wait time after PS detection	7:0	0 to 254	2.5 ms x (PS_WAIT + 1)	

PS\_WAIT defines the wait time between each detection. In most applications, the wait time between 2.5 ms and 637.5 ms is enough to find a compromise between acceptable performance and low power consumption. A higher measurement rate (shorter wait time) increases the amount of data and decreases the possibility of false detection. However, the sensor will consume more current. The designer could select shorter wait time options in PS\_WAIT if a high measurement rate is required. Fig. 19 and Fig. 20 show examples of the wait time setting.





#### 4.5 Offset Count Cancellation

TABLE 12	TABLE 12 - OFFSET COUNT CANCELLATION								
REGISTER GROUP	BIT NAME	REGISTER NAME	COMMAND CODE	FUNCTION	BIT	VALUE	DESCRIPTION		
Offset count	PS_CANC_L		0×06	Set the offset count	7:0	0 to 4005	Low byte		
cancellation	PS_CANC_H	F3_CANC	0200	cancellation value	11:8	0 10 4095	High byte		

Crosstalk between the infrared signal from the VCSEL and the photodiode happens when introducing the window cover. This is because part of the light will be reflected back to the sensor due to fresnel reflection. Therefore, it is recommended to perform offset count cancellation by writing the PS\_CANC register with the offset counts. Fig. 21 shows the offset count calibration by measuring the offset counts due to window cover, internal crosstalk, and noise. These offset counts can then be written to the register PS\_CANC, as shown in Fig. 22. By using this approach, the change in counts is directly influenced by the object and the effect of the part to part tolerance can be reduced.



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Fig. 21 - Offset Count Calibration

#### 4.6 Sunlight Cancellation

TABLE 1	TABLE 13 - SUNLIGHT CANCELLATION									
REGISTER GROUP	BIT NAME	REGISTER NAME	COMMAND CODE	FUNCTION	BIT	VALUE	DESCRIPTION			
	PS_SC_LEVEL					0x0 (0b00)	Level 1 (default)			
		PS_CONF4_H	0x03	Sunlight cancellation level setting cancellation value	11 . 10	0x1 (0b01)	Level 2			
Sunlight					11.10	0x2 (0b10)	Level 3			
cancellation						0x3 (0b11)	Level 4			
				Enable / disable	0	0x0 (0b0)	Disable (default)			
	F3_50			the sunlight cancellation	9	0x1 (0b1)	Enable			

DC ambient light sources in the wavelength region between 800 nm and 1200 nm, such as sunlight and halogen, cause disturbances in the photodiode and the ADC circuitry. Therefore, the sensor needs to perform cancellation to reduce the noise from these disturbances. The sunlight cancellation PS\_SC bit can be enabled to allow the sensor to measure the photocurrent contribution from the DC ambient light sources before driving the infrared signal pulse during the proximity measurement. After the proximity measurement, the sensor deducts the DC noise photocurrent from the total photocurrent. As a result, only the photocurrent due to the reflected signal from the object is converted by the ADC circuitry. This sunlight cancellation mechanism is depicted in Fig. 23.



Fig. 23 - Sunlight Cancellation by Using the Active Sunlight Cancellation Current in the VCNL36829UM



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The sensor can perform the cancellation of sunlight up to 250 klux. The sensor then goes into the sunlight protection mode beyond 250 klux of sunlight. Sunlight protection mode is a mode in which the photodiode is in saturation due to high sunlight illuminance. Therefore, the sensor can no longer detect the object and will only output 0 in the PS\_DATA register. This can be observed in Fig. 24. If the sunlight protection mode interrupt PS\_SP\_INT is enabled, the sensor will pull the interrupt line low each time the sensor goes into sunlight protection mode. As a result, the sunlight protection mode interrupt flag PS\_SPFLAG changes from 0 to 1. This interrupt flag can be cleared by reading the INT\_FLAG register. Therefore, the application should ignore the PS\_DATA when it remains 0 and PS\_SPFLAG consistently changes to 1 after clearing the INT\_FLAG register.



Fig. 24 - The Overall Behavior of the Sunlight Cancellation in the Proximity Sensor With Increasing Sunlight Illuminance

If the PS\_SC bit is disabled, the sensor goes into the sunlight protection mode when the illuminance of the sunlight is beyond 10 klux. This can be observed in Fig. 24. When PS\_SC\_EN is set to 1, PS\_SC\_LEVEL can be used to select different sunlight cancellation levels. The Table 14 below shows the capability of each sunlight cancellation setting.

TABLE 14 - THE MAXIMUM SUNLIGHT LUX LOAD FOR SUN_LEVEL						
Disable	10K					
Level 1	90K					
Level 2	140K					
Level 3	180K					
Level 4	250K					



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#### 4.7 Interrupt

TABLE	TABLE 15 - INTERRUPT									
REGISTER GROUP	BIT NAME	REGISTER NAME	COMMAND CODE	FUNCTION	BIT	VALUE	DESCRIPTION			
						0x0 (0b00)	Interrupt disable (default)			
				Set the interrupt	15.14	0x1 (0b01)	Logic high / low mode			
	F3_IN1			mode setting	15.14	0x2 (0b10)	First high			
						0x3 (0b11)	Trigger by each high / low threshold event			
				Set the amount of		0x0 (0b00)	1 time (default)			
	PS PERS			consecutive threshold	13 . 12	0x1 (0b01)	2 times			
	10_1 110	PS CONF1 H	0x00	crossing events necessary to	10.12	0x2 (0b10)	3 times			
				inggerimenupi		0x3 (0b11)	4 times			
				Enable / disable the smart		0x0 (0b0)	Disable (default)			
	PS_SMART_PERS			interrupt event is triggered	11	0x1 (0b1)	Enable			
				Enable / disable the		0x0 (0b0)	Disable (default)			
	PS_SP_INT			sunlight protection mode interrupt setting	10	0x1 (0b1)	Enable			
				After PS_SD disable,	_	0x0 (0b0)	Disable (default)			
	PS_START_INT			PS 1 <sup>st</sup> detection interrupt enable setting	9	0x1 (0b1)	Enable			
	PS_THDL_L		0×04	Set the low threshold	7:0	0 to 65535	Low byte			
Interrupt	PS_THDL_H	13_IIIDE	0,04	interrupt value	15 : 8	01000000	High byte			
	PS_THDH_L	PS_THDH_L PS_THDH		Set the high threshold	7:0	0 to 65535	Low byte			
	PS_THDH_H	10_111011	0,000	interrupt value	15 : 8	0.000000	High byte			
				PS finish 1st detection		0x0 (0b0)	No 1st detection flag			
	PS_START_FALG			after SD disable	11	0x1 (0b1)	Finish 1 <sup>st</sup> detection flag			
				Read the sunlight protection	10	0x0 (0b0)	No sunlight protection mode flag			
	10_011EAG			mode interrupt event flag	10	0x1 (0b1)	Sunlight protection mode flag			
		INT_FLAG	0xF9	Read the high threshold	٩	0x0 (0b0)	No high threshold crossing interrupt event flag			
	F3_II_0L03L			crossing interrupt event flag	5	0x1 (0b1)	High threshold crossing interrupt event flag			
				Read the low threshold	8	0x0 (0b0)	No low threshold crossing interrupt event flag			
				crossing interrupt event flag	0	0x1 (0b1)	Low threshold crossing interrupt event flag			



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The interrupt pin allows the proximity sensor to autonomously send an interrupt signal to the microcontroller when the sensor reading exceeds the high threshold or falls below the low threshold. There are three modes of interrupt in the VCNL36829UM:

- Logic high / low mode
- First high
- Trigger by each high / low threshold event

Besides that, the proximity sensor also provides persistence and smart persistence features, which prevent the occurrence of false detection. The interrupt will only be triggered after a defined consecutive threshold-crossing event.

#### Logic High / Low Mode

- The interrupt line is pulled low when the proximity counts cross the high threshold, as indicated by the blue line in Fig. 25
- The interrupt line is pulled high when the proximity counts cross the low threshold, as indicated by the purple line in Fig. 25
- If the count is equal to the threshold, it cannot be triggered, as shown by the orange circles in Fig. 25
- Consecutive high threshold events cannot be triggered until the proximity count crosses the low threshold first, and vice versa, as shown by the green circles in Fig. 25
- Logic mode allows the proximity sensor to autonomously send an interrupt signal to the microcontroller directly without the need to read the INT\_FLAG register



Fig. 25 - The Behavior of the INT Line (pin 1) Using Logic High / Low Mode With PS\_PERS = 1



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### Trigger by Each High / Low Threshold Event

- The interrupt line is pulled low when the proximity counts cross the high or low threshold, as indicated by the red circle in Fig. 26
- If the count is equal to the threshold, it cannot be triggered, as shown by the orange circles in Fig. 26
- The first interrupt event can be either a high or low threshold crossing event
- The interrupt flags PS\_IF\_CLOSE and PS\_IF\_AWAY change from 0 to 1 when the proximity counts cross the high and low thresholds, respectively
- The interrupt flag PS\_IF\_CLOSE or PS\_IF\_AWAY can be cleared by reading the INT\_FLAG register, as indicated by the blue line in Fig. 26



Fig. 26 - The Behavior of the INT Line (pin 1) and INT\_FLAG Register Using the Trigger by Each High / Low Threshold Event With PS\_PERS = 1

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### First High

- The interrupt line is pulled low when the proximity counts cross the high or low threshold, as indicated by the red circle in Fig. 27
- Consecutive events can be triggered until the proximity count crosses the threshold first, as shown by the green circles in Fig. 27
- If the count is equal to the threshold, it cannot be triggered, as shown by the orange circles in Fig. 27
- The first interrupt event can be either a high or low threshold crossing event
- The interrupt flags PS\_IF\_CLOSE and PS\_IF\_AWAY change from 0 to 1 when the proximity counts cross the high and low thresholds, respectively
- The interrupt flag PS\_IF\_CLOSE or PS\_IF\_AWAY can be cleared by reading the INT\_FLAG register, as indicated by the blue line in Fig. 27



#### Persistence and Smart Persistence Mode

PS\_PERS determines the number of consecutive threshold-crossing events required to trigger an interrupt. When PS\_PERS is set between 2 and 4, three additional measurements are taken after the initial threshold crossing event to confirm the persistence of the crossing, ensuring false detections are minimized. The interrupt is only triggered after the number of consecutive threshold-crossing events, as specified by the PS\_PERS setting, has been reached. This applies to all interrupt modes. Fig. 28 illustrates the behavior of the INT line (pin 1) with a measurement period as defined by the red line formula, where the period equals PS\_WAIT + DC\_KILL + (2 × PS\_IT). Fig. 28 also shows how the INT line performs under different PS\_PERS settings. For example, when PS\_PERS is set to 2, the interrupt is pulled low after the second threshold-crossing measurement out of the four measurements following the initial threshold-crossing event. The same logic applies to other PS\_PERS settings.



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When smart persistence (PS\_SMART\_PERS) is enabled after the first threshold event occurs, three consecutive rapid measurements are taken. After the third measurement, the system resumes its regular measurement period. The intervals between these measurements are depicted by the green line in the figure, and the interval is 1.4 ms. Fig. 29 shows the behavior of the INT line (pin 1), where the measurement period follows the red line formula in the figure, with the period equal to  $PS_WAIT + DC_KILL + (2 \times PS_IT)$ . Fig. 29 also illustrates how the INT line performs under different PS\_PERS settings. For example, when PS\_PERS is set to 3, the interrupt is pulled low after the third consecutive rapid measurement that crosses the threshold. The same logic applies to other PS\_PERS settings.



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#### **PS INT Response Time**

Disabled PS\_SMART\_PERS:

Response time (ms) =  $PS_PERS \times (PS_WAIT + DC_KILL + (2 \times PS_IT))$ 

Enabled PS\_SMART\_PERS:

Response time (ms) =  $(PS_PERS - 1) \times (1.4 + \text{total time}) + (PS_WAIT + DC_KILL + (2 \times PS_IT))$ Total time = DC\_KILL + (2 × PS\_IT)

#### **Sunlight Protection Mode**

The sensor enters the sunlight protection mode when the photodiode is in saturation due to very high sunlight illuminance. Therefore, the sensor can no longer detect the object and will only output 0 in the PS\_DATA register. The application can detect this situation by enabling PS\_SP\_INT and reading the PS\_SPFLAG when the interrupt line has been pulled low and PS\_DATA suddenly goes into 0 to confirm this situation.

#### **PS Start Interrupt**

- The interrupt line is pulled low when finished with the first detection after PS\_SD disable, as indicated by the blue line in Fig. 30
- The interrupt flag PS\_START\_FLAG can be cleared by reading the INT\_FLAG register, as indicated by the purple line in Fig. 30



Fig. 30 - The Behavior of the INT Line (pin 1), Enabled PS\_START\_INT and INT\_FLAG Register



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#### 4.8 Readout Registers

TABLE 16 - READOUT REGISTERS								
REGISTER GROUP	BIT NAME	REGISTER COMMAND NAME CODE FUNCTION		BIT	VALUE	DESCRIPTION		
	PS_DATA_L			Read the proximity	7:0	0 to 65 535	Low byte	
	PS_DATA_H	FS_DATA	0110	output data	15 : 8	0 10 03 333	High byte	
Readout	VCNL36829UM_ID_L				7:0	0x29 (0b00101001)	Should be kept default	
registers		VCNL36829UM_ID	0xFA	Read the device ID	15.0	0x00 (0b0000000)	Device with a slave address of 0x60	
					15.0	0x10 (0b00010000)	Device with a slave address of 0x51	

The VCNL36829UM has two readout registers, which are PS\_DATA and VCNL36829UM\_ID. PS\_DATA is the 16-bit register in which the proximity counts data will be stored after each measurement. The register can be read via the I<sup>2</sup>C communication from a microcontroller. There are two methods of reading the register PS\_DATA:

- Data polling continuously reading the register PS\_DATA
- Interrupt read the register PS\_DATA after the INT line (pin 1) has been triggered

Depending on the application's requirements, the interrupt method is recommended to reduce power consumption.

On the other hand, VCNL36829UM\_ID is the register in which the device ID is stored. This allows the sensor to be identified with a specific slave address depending on the connection of pins 5 and 6 of the sensor with the microcontroller. Therefore, the register VCNL36829UM\_ID can be a good first register to read when first communicating with the sensor to test the I<sup>2</sup>C communication and to check the slave address.





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### 5. I<sup>2</sup>C AND TIMING

#### 5.1 I<sup>2</sup>C Write and Read Protocol

The communication with the VCNL36829UM can be performed via I<sup>2</sup>C. The I<sup>2</sup>C write and read protocol when communicating with the proximity sensor is shown in Fig. 31.

Send byte → write command to VCNL36829UM																
1	7	1	1	8	1		8	1			8 1	1				
S	Slave address	Wr	А	Command code	А		Data byte low	А	Data byte high		a byte high A	Ρ				
Rece	eceive byte $\rightarrow$ read data from VCNL36829UM															
1	7	1	1	8	1	1	7		1	1	8		1	8	1	1
s	Slave address	Wr	А	Command code	А	S	Slave address		Rd	А	A Data byte low		А	Data byte high	Ν	Р
S = start condition Host action   P = stop condition Host action   A = acknowledge VCNL36829UM response				e												

Fig. 31 - I<sup>2</sup>C Write and Read Protocol

It is imperative that only the restart condition for the I<sup>2</sup>C read is implemented instead of the stop and restart condition. For example, Fig. 32 shows a logic analyzer screenshot of the SCL (pin 5) and SDA line (pin 6) when reading the register VCNL36829UM\_ID for the VCNL36829UM with a slave address of 0x60. Here, the restart condition, indicated by the blue box, has been implemented and the low data byte of 0x29 matches the device ID stated in the datasheet for the VCNL36829UM with the slave address of 0x60.



Fig. 32 - The Correct I<sup>2</sup>C Read When Reading the Register VCNL36829UM\_ID With a Restart Condition

#### 5.2 Timing

#### 5.2.1 Proximity Measurement Timing

One active measurement period for both depends on the register settings PS\_IT, as well as DC\_KILL, as shown below:  $t_{measurement, active} = 1.2 \times (DC_KILL + (2 \times PS_IT))$ 

Factor 1.2 is included in the equation above to consider the part to part tolerance of 20 %. On the other hand, the term DC\_KILL is the process where the sensor measures the DC noise in the background before the actual measurement pulse from the VCSEL is emitted. This DC noise signal will then be deducted from the total reflected signal.

#### **Example - Active Measurement Time**

For example, given PS\_IT = 25  $\mu$ s, the total active measurement time can be calculated as follows:

 $t_{measurement, active} = 1.2 \text{ x} (278.6 \ \mu\text{s} + (2 \ \text{x} \ 25 \ \mu\text{s})) = 394 \ \mu\text{s}$ 

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#### 5.2.2 Timing Specification

Fig. 33 and Fig. 34 show the timing specification for the auto and active force modes, respectively.



Fig. 33 - Timing Specification for the Auto Mode

#### Notes

- (1) B µs The parameter B refers to the time taken for a complete write I<sup>2</sup>C protocol. This depends on the selected I<sup>2</sup>C mode. Please refer to Table 17
- (2) D µs The parameter D refers to the time taken for a complete read I<sup>2</sup>C protocol. This depends on the selected I<sup>2</sup>C mode. Please refer to Table 17



Fig. 34 - Timing Specification for the Active Force Mode

#### Notes

- (1) B µs The parameter B refers to the time taken for a complete write I<sup>2</sup>C protocol. This depends on the selected I<sup>2</sup>C mode. Please refer to Table 17
- (2) D µs The parameter D refers to the time taken for a complete read I<sup>2</sup>C protocol. This depends on the selected I<sup>2</sup>C mode. Please refer to Table 17

A measurement delay equal to the active measurement time is required for the measurement to be completed after the trigger of the active force mode. The delay depends on the register settings of PS\_IT. The required delay is:

Measurement delay =  $t_{measurement, active} = (1.2 \text{ x DC}_KILL + (2 \text{ x PS}_IT))$ 

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Factor 1.2 in the equation is needed to consider the 20 % part to part tolerance of the measurement. The individual  $I^2C$  write and read time per byte depends on the  $I^2C$  mode, as shown in Table 17:

TABLE 17 - I <sup>2</sup> C WRITE AND READ PROTOCOL TIME PER BYTE							
I <sup>2</sup> C MODE	CLOCK FREQUENY (kHz) <sup>(1)</sup>	I <sup>2</sup> C WRITE PROTOCOL TIME (µs) <sup>(2)</sup>	I <sup>2</sup> C READ PROTOCOL TIME (µs) <sup>(2)</sup>				
Standard	100	400	500				
Fast	400	100	125				

Notes

(1) Maximum limit

<sup>(2)</sup> Typical value with tolerance; it could vary depending on master

#### Example - Auto Mode With Standard I<sup>2</sup>C Mode

Table 18 shows an example of the timing specification for the auto mode when initializing seven registers, as well as when turning on the sensor. The sensor needs approximately 5.7 ms for the total initialization phase. The exact timing depends on the I<sup>2</sup>C clock configuration from the master and the number of registers written. On the other hand, the sensor needs one measurement period to complete a measurement. In the case of when a PS\_WAIT of 2.5 ms and PS\_IT of 100 µs has been selected, the measurement time is 3 ms. In many cases, the application uses the interrupt method to read the sensor and reduce power consumption. In this case, once the INT line (pin 1) has been pulled low, the interrupt service routine in the microcontroller will be executed and the microcontroller will then read the register INT\_FLAG to clear the interrupt flag. The register PS\_DATA will then be read. Both reading processes will each need 0.5 ms for the standard I<sup>2</sup>C mode. Therefore, the total measurement time is 4 ms.

TABLE 18 - TIMING SPECIFICATION EXAMPLE FOR AUTO MODE AND STANDARD I <sup>2</sup> C MODE								
PARAMETER	REGISTER	REMARKS	TIME (ms)					
t <sub>on</sub>	-	The power on reset after the $V_{DD}$ pin is connected to the power supply	2.5					
	PS_CONF1_L/H		0.4					
t <sub>initialization</sub>	PS_CONF2_L/H		0.4					
	PS_CONF3_L/H	Write the sensor settings	0.4					
	PS_CONF4_L/H		0.4					
	PS_CONF5_L/H		0.4					
	PS_THDL_L/H	/H Write the low threshold setting						
	PS_THDH_L/H	Write the high threshold setting	0.4					
	PS_CONF1_L/H	Turn on the sensor by writing PS_SD = 0	0.4					
t <sub>initialization_total</sub>	-	-	5.7					
t <sub>measurement</sub>	-	Actual measurement time depends on the selected PS_WAIT and PS_IT; here is an example of when PS_WAIT = $2.5 \text{ ms}$ , PS_IT = $100 \mu s$	3					
+	INT_FLAG	Read the interrupt flag	0.5					
<sup>L</sup> read	PS_DATA	Read the proximity data	0.5					
t <sub>measurement_total</sub>	-	-	4					

#### Notes

(1) Standard I<sup>2</sup>C mode comes with a 100 kHz clock frequency. One can approximate the time taken for a complete write and read I<sup>2</sup>C protocol to be 400 µs and 500 µs per byte, respectively. Please refer to Table 17

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### 6. PROXIMITY MEASUREMENT

#### 6.1 Software Flow Chart for the Proximity Measurement

Fig. 35 shows a typical software flow chart for the proximity measurement using the auto mode and enabled interrupt. The measurement starts with the initialization code. Please note that the provided sensor initialization code is just an example. In practice, the designer should change the initialization code based on the application's requirements. After the initialization, the sensor should perform the offset count cancellation by reading the proximity count without the object. This offset count should then be written to the offset cancellation register PS\_CANC. After writing the cancellation register, the sensor should perform proximity measurement in a loop. Next, the microcontroller should wait for the interrupt line to be pulled low via an interrupt service routine. When the interrupt has been pulled low, the interrupt flag register and the proximity data will be read.



Fig. 35 - Typical Software Flow Chart for the Auto Mode

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#### 6.2 Proximity Value vs. Distance Curve

The behavior of the proximity sensor can be characterized by the proximity value against the distance. In some applications, the sensor will not be placed underneath a window cover. The designer could then use the proximity value against the distance for the case without the window cover as a reference. On the other hand, most applications will have the requirement of placing the sensor underneath a window cover. Here, the proximity value against the distance for the case with the window cover can be used as a reference. The designer can expect a decrease in the dynamic range due to the increase of crosstalk from the window cover.

#### 6.2.1 Proximity Value vs. Distance Curve Without a Window Cover

Fig. 36 and Fig. 37 show the typical proximity counts output against distance for three different driving currents and two integration times. Here, the measured reference object medium is the Kodak Gray Card. This card has approximately 18 % reflectivity at 940 nm. The sensor can detect up to 5 cm with a minimum of 960 delta counts when measuring with the Kodak Gray Card and when higher sensor settings of PS\_IT = 200  $\mu$ s and PS\_CURRENT = 18 mA have been selected. The dashed red line indicates this in Fig. 37. The minimum of 960 delta counts is usually enough to consider the sensor's tolerance of  $\pm$  20 % and account for the noise from the disturbing sources.



Fig. 36 - Proximity Value vs. Distance; PS\_IT = 50 µs (14 bit)

Fig. 37 - Proximity Value vs. Distance; PS\_IT = 200 µs (16 bit)

While the Kodak Gray Card is used as a reference, in practice, the proximity values change with the different objects' reflectivity at the same distance. Therefore, the designer should consider this effect during the design phase.





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#### 6.3 Temperature Behavior

When designing the VCNL36829UM for the end application, the temperature behavior of the proximity counts should also be considered. Fig. 38 shows the average relative proximity counts against ambient temperature of the VCNL36829UM.





The sensor has a negative temperature coefficient as the temperature increases from room temperature to 85 °C. The designer should consider this behavior and design the end application with enough headroom for the proximity count's fluctuation with temperature.

Another good design approach to reduce the effect of the proximity count's fluctuation due to temperature is regular calibration without an object. One can observe that the change in proximity counts with temperature happens slower than the change in proximity counts with changing distance of an object. By observing the increase or decrease of offset counts, the system can recognize that as the change of temperature and proximity correction could be implemented.



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### 7. AVERAGE CURRENT CONSUMPTION

The VCNL36829UM proximity sensor has been developed for battery-powered devices such as true wireless stereo (TWS) earbuds, smart watches, etc. The designer needs to compromise between the performance and current consumption for such applications.

Fig. 39 shows the current consumption components through the pins.



Fig. 39 - Current Consumption Components Through the Pins

The general total current consumption can be calculated as follows:

$$I_{\text{total}} = I_{\text{DD}} + I_{\text{VCSEL}} + I_{1^2 \text{C}}$$

The main current components are from  $I_{DD}$  and  $I_{VCSEL}$ . The current contribution from  $I_{I^2C}$  for  $I^2C$  communication is negligible if the interrupt is activated. However, if the polling mode is used, the current contribution from these pins should be considered in the calculation.

Fig. 40 shows the current consumption profile of the VCNL36829UM.





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Table 19 describes the parameters of the current consumption profile in Fig. 40.

TABLE 19	TABLE 19 - AVERAGE CURRENT CONSUMPTION PARAMETER							
PARAMETER	PARAMETER NAME	VALUE	DESCRIPTION					
А	I <sub>VCSEL</sub>	8 mA to 18 mA	The magnitude of $I_{VCSEL}$ depends on the selected PS_CURRENT setting; this current is drawn by the pin VCSEL_A					
В	I <sub>DD</sub> (Active)	310 µA	$I_{\text{DD}}$ (active) is the supply current consumed during the active state by the pin $V_{\text{DD}}$					
С	I <sub>DD</sub> (Idle)	5 µA	$I_{\text{DD}}$ (idle) is the supply current consumed during the idle state by the pin $V_{\text{DD}}$					
D	DC_KILL	278.6 µs	DC_KILL is the period where the sunlight and DC noise cancellation happens					
E	Proximity measurement pulse	2 x PS_IT	Proximity measurement is the duration in which the sensor sends the VCSEL pulse/s; the values depend on PS_IT					
D + E <sup>(1)</sup>	- E <sup>(1)</sup> Active DC_KILL + measurement period (2 x PS_IT)		The active measurement period is the duration during which the sensor performs the complete proximity measurement, which consists of DC_KILL and proximity measurement pulse					
F	Measurement period DC_KILL + (2 > PS_IT) + PS_WA		The measurement period in which the measurement is repeated, which depends on the selected PS_IT and PS_WAIT setting					

#### Note

(1) The active measurement period could have ± 20 % part to part tolerance, which is not considered here

A low current consumption can be achieved in the VCNL36829UM due to a low supply current of 5 µA during the idle state. When the sensor is switched on, it will always consume 5 µA. This is indicated by the parameter "C" in Fig. 40 over the whole time, as long as the sensor is switched on. During the active measurement period, as indicated by the parameters "D" and "E", the sensor consumes an additional current of 310 µA on top of 5 µA. Besides that, during the active measurement period, the sensor will emit a short current pulse PS\_CURRENT with a duration of PS\_IT.

The average current consumption over the measurement period in the VCNL36829UM with interrupt mode can be calculated as follows:

$$I_{average} = I_{DD}(idle) + \frac{((DC_KILL + (2 \times PS_IT)) \times I_{DD}(active)) + (PS_IT \times PS_CURRENT)}{DC_KILL + (2 \times PS_IT) + PS_WAIT}$$

#### **Example - Average Current Consumption Calculation**

Table 20 shows the average current consumption example for PS\_IT = 25 µs for different PS\_WAIT and PS\_CURRENT.

PS_IT	PS_CURRENT (mA)	PS_WAIT (ms)													
(µs)		2.5	50	100	150	200	250	300	350	400	450	500	550	600	637.5
25	8	111.7	11.0	8.0	7.0	6.5	6.2	6.0	5.9	5.8	5.7	5.6	5.5	5.5	5.5
25	10	129.4	12.0	8.5	7.3	6.8	6.4	6.2	6.0	5.9	5.8	5.7	5.6	5.6	5.6
25	12	147.1	13.0	9.0	7.7	7.0	6.6	6.3	6.1	6.0	5.9	5.8	5.7	5.7	5.6
25	14	164.7	14.0	9.5	8.0	7.3	6.8	6.5	6.3	6.1	6.0	5.9	5.8	5.8	5.7
25	16	182.4	15.0	10.0	8.3	7.5	7.0	6.7	6.4	6.3	6.1	6.0	5.9	5.8	5.8
05	18	200.1	16.0	10.5	8.7	7.8	7.2	6.8	6.6	6.4	6.2	6.1	6.0	5.9	5.9

$$I_{\text{average}} = 5 \ \mu\text{A} + \frac{((278.6 \ \mu\text{s} + (2 \ \text{x} \ 25 \ \mu\text{s})) \ \text{x} \ 310 \ \mu\text{A}) + (25 \ \mu\text{s} \ \text{x} \ 8 \ \text{mA})}{278.6 \ \mu\text{s} + (2 \ \text{x} \ 25 \ \mu\text{s}) + 100 \ \text{ms}} = 8 \ \mu\text{A}$$

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Table 21 shows the average current consumption example for PS\_IT = 200 µs for different PS\_WAIT and PS\_CURRENT.

TABLE 21 - AVERAGE CURRENT CONSUMPTION															
PS_IT	PS_CURRENT (mA)	PS_WAIT (ms)													
(µs)		2.5	50	100	150	200	250	300	350	400	450	500	550	600	637.5
200	8	574.5	40.7	23.0	17.0	14.0	12.2	11.0	10.2	9.5	9.0	8.6	8.3	8.0	7.8
200	10	700.4	48.6	27.0	19.7	16.0	13.8	12.4	11.3	10.5	9.9	9.4	9.0	8.7	8.5
200	12	826.2	56.5	30.9	22.3	18.0	15.4	13.7	12.4	11.5	10.8	10.2	9.7	9.3	9.1
200	14	952.1	64.4	34.9	25.0	20.0	17.0	15.0	13.6	12.5	11.7	11.0	10.5	10.0	9.7
200	16	1077.9	72.3	38.9	27.6	22.0	18.6	16.3	14.7	13.5	12.6	11.8	11.2	10.7	10.3
200	18	1203.8	80.2	42.8	30.3	24.0	20.2	17.7	15.9	14.5	13.5	12.6	11.9	11.3	11.0

### 8. MECHANICAL AND OPTICAL DESIGN

As light waves propagate through space, they naturally spread out. This phenomenon also occurs with the infrared light emitted by the VCSEL, which broadens as it moves farther from the sensor. Consequently, the diameter of the window cover must be carefully adjusted to meet the minimum size requirements, ensuring that the opaque housing does not obstruct the infrared signal. This minimum diameter is influenced by the distance between the top of the cover and the top of the sensor, the emission angle of the VCSEL, and the photodiode's field of view.

Typically, the window diameter needs to increase as the distance from the sensor increases. The VCSEL in the sensor emits light with an angle of  $\pm 4.5^{\circ}$ . Conversely, the photodiode in the sensor has a half-sensitivity angle of  $\pm 60^{\circ}$  in the X-axis or the Y-axis direction. These angles are illustrated in Fig. 41 and Fig. 42.







Fig. 42 - Relative Sensitivity vs. Angular Displacement of the Photodiode in the VCNL36829UM

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Fig. 43 - The Calculation of the Window Cover Diameter With an Air Gap

#### **Explanation of Variables**

- x: this represents the distance between the center of the VCSEL and the center of the photodiode (PD). In Fig. 43, this is the horizontal distance between the red and orange blocks. The value of x is given as 0.85 mm, which includes the mechanical placement tolerance of the VCSEL and photodiode within the sensor package
- y: this is the distance from the top of the window cover to the top of the sensor. It represents the vertical distance between the black line at the top (which marks the position of the window cover) and the top of the sensor. This is an important parameter, as it determines the necessary window cover size to avoid blocking the infrared signal
- θ1: this angle represents the emission angle of the VCSEL. It is the angle between the red dashed line (indicating the direction of the emitted signal) and the vertical axis. Fig. 43 suggests that θ1 should ideally be 4.5°, but a larger angle (e.g., 10°) may be used to account for mechanical tolerance and ensure proper signal transmission
- $\theta$ 2: this angle represents the field of view of the photodiode (PD). It is the angle between the orange dashed line (indicating the field of view) and the vertical axis. Fig. 43 shows this as an angle that allows the photodiode to capture the reflected infrared signal from the object. The recommended value is  $\theta$ 2 = 50°, though the photodiode's total field of view is ± 60°
- d: represents the distance between the edge of the package near the photodiode (PD) and the window cover. This distance assists with alignment during assembly

#### **Calculation of Minimum Window Cover Diameter D**

The minimum window cover diameter D can be calculated using the following formulas:

$$\mathsf{D} = \mathsf{y} \times \tan \theta_2 + \mathsf{x} + \mathsf{y} \times \tan \theta_1$$

This formula considers both the emission angle  $\theta$ 1 and the field of view  $\theta$ 2 to determine the necessary window cover diameter, ensuring that the infrared signal is not blocked during either emission or reception.





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#### Calculation of Minimum Distance d

The distance between the photodiode (PD) and the edge of the package can be calculated using the following formula:

 $d = 0.48 - y \times tan\theta_2$ 

When y exceeds 0.4 mm, d becomes negative, indicating that the edge of the window will be on the right side of the package. Conversely, if d is positive, the edge of the window will be on the left side of the package.

#### **Additional Considerations**

The placement of the window cover directly on top of the sensor is usually preferred, as this avoids the sensor's peak sensitivity area and helps reduce crosstalk. However, achieving this during assembly can be challenging due to mechanical tolerance, which often results in an air gap. If an air gap is present, it should ideally be less than 1 mm to minimize its impact on the sensor's performance.

TABLE 22 - MINIMUM WINDOW COVER DIAMETER								
y (mm)	D (mm)	d (mm)						
0.5	1.53	0.12						
1.0	2.22	0.71						
1.5	2.90	1.31						
2.0	3.59	1.90						
2.5	4.27	2.50						
3.0	4.95	3.10						
3.5	5.64	3.69						
4.0	6.32	4.29						