



# Designing the VCNL3040 Into an Application

By Reinhard Schaar

## INTRODUCTION AND BASIC OPERATION

The VCNL3040 is a fully integrated proximity sensor. It combines an infrared emitter and photodiode for proximity measurement and signal processing IC in a single package with a 12-bit / 16-bit ADC. The device provides proximity sensing to minimize accidental touch input that can lead to call drops and camera launch.

With a range of up to 30 cm (12"), this stand-alone component greatly simplifies the use and design-in of a proximity sensor in consumer and industrial applications, because no mechanical barriers are required to optically isolate the emitter from the detector. The VCNL3040 features a miniature, surface-mount 4.0 mm by 2.0 mm leadless package (LLP) with a low profile of 1.1 mm. The device is designed specifically to meet the low height requirements of smartphone, mobile phone, digital camera, and tablet PC applications.

Through its standard I<sup>2</sup>C bus serial digital interface, it allows easy access to a "proximity signal". The programmable interrupt function offers wake-up functionality for the microcontroller when a proximity event occurs, which reduces processing overhead by eliminating the need for continuous polling.



Fig. 1 - VCNL3040 Top View

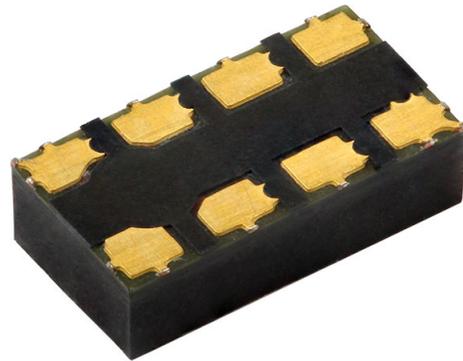


Fig. 2 - VCNL3040 Bottom View

## COMPONENTS (BLOCK DIAGRAM)

The major components of the VCNL3040 are shown in the block diagram.

In addition to the ASIC with the proximity photodiode, the infrared emitter is also implemented. Its cathode is connected to the driver internally and needs not to be connected externally.

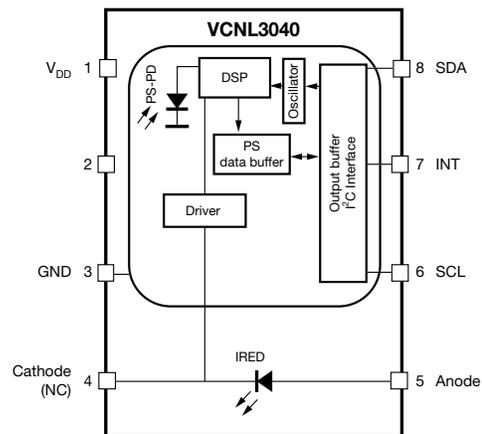


Fig. 3 - VCNL3040 Detailed Block Diagram

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The integrated infrared emitter has a peak wavelength of 940 nm. It emits light that reflects off an object within 30 cm of the sensor. An added lens helps to increase peak intensity due to enabling a small angle of just  $\pm 15^\circ$ , as shown in Fig. 6.

The infrared emitter has a programmable drive current from 50 mA to 200 mA in eight steps. The infrared light is emitted in short pulses with a programmable duty ratio from 1/40 to 1/320. The proximity photodiode receives the light that is reflected off the object and converts it to a current. The sensitivity of the proximity stage is also programmable by choosing from eight different integration times. Due to the wider spectral response it is more sensitive to ambient light, especially direct sunlight.

The application-specific integrated circuit, or ASIC, includes an LED driver, I<sup>2</sup>C bus interface, amplifier, integrated analog-to-digital converter, oscillator, and Vishay's "secret sauce" signal processor. For proximity, it converts the current from the photodiode to a 12-bit or 16-bit digital data output value.

### PIN CONNECTIONS

Fig. 3 shows the pin assignments of the VCNL3040.

The connections include:

- Pin 1 - V<sub>DD</sub> to the power supply
- Pin 2 - no connection
- Pin 3 - connect to ground
- Pin 4 - IRED cathode (no connection)
- Pin 5 - IRED anode to the power supply
- Pin 6 - SCL to microcontroller
- Pin 7 - INT to microcontroller
- Pin 8 - SDA to microcontroller

The power supply for the ASIC (V<sub>DD</sub>) has a defined range from 2.5 V to 3.6 V. The infrared emitter can also be within this range and also here 3.6 V is maximum. It is best if V<sub>DD</sub> is connected to a regulated power supply and pin 5, the anode, is connected directly to the battery. This eliminates any influence of the high infrared emitter current pulses on the V<sub>DD</sub> supply line. The power supply decoupling components C1 and R4 (shown in Fig. 4) are optional. They isolate the sensor from other possible noise on the same power rail, but in most applications are not needed.

If separate power supplies for the V<sub>DD</sub> and the infrared emitter are used and there are no negative spikes below 2.5 V, only one capacitor at V<sub>DD</sub> could be used. The 100 nF capacitor should be placed close to the V<sub>DD</sub> pin. The SCL and SDA, as well as the interrupt lines, need pull-up resistors. The resistor values depend on the application and on the I<sup>2</sup>C bus speed. Common values are about 2.2 k $\Omega$  to 4.7 k $\Omega$  for the SDA and SCL, and about 8.2 k $\Omega$  to 22 k $\Omega$  for the interrupt lines.

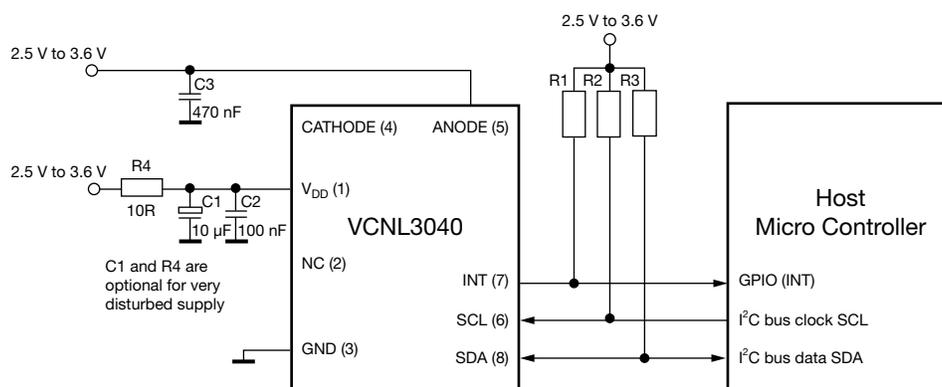


Fig. 4 - VCNL3040 Application Circuit

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### MECHANICAL DESIGN CONSIDERATIONS

The VCNL3040 is a fully integrated proximity sensor. Competing sensors use a discrete infrared emitter, which leads to complex geometrical calculations to determine the position of the emitter. Competing sensors also require a mechanical barrier between the emitter and detectors to eliminate crosstalk - light reflecting off the inside of the window cover that can produce false proximity readings.

The VCNL3040 does not require a mechanical barrier. The signal processor continuously compensates for the light reflected from windows, thus ensuring a proper proximity reading. As a fully integrated sensor, the design process is greatly simplified.

The only dimensions that the design engineer needs to consider are the distance from the top surface of the sensor to the outside surface of the window, and the size of the window. These dimensions will determine the size of the detection zone.

The angle of half intensity of the emitter is about  $\pm 15^\circ$ , as shown in Fig. 5, and the sensitivity of the photodiode is showing about  $\pm 55^\circ$ .

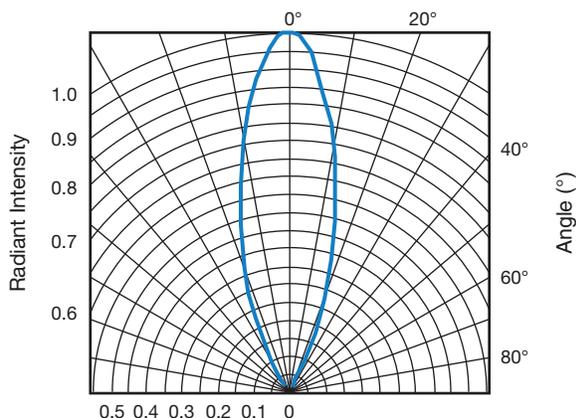


Fig. 5 - Angle of the Half Intensity of the Emitter

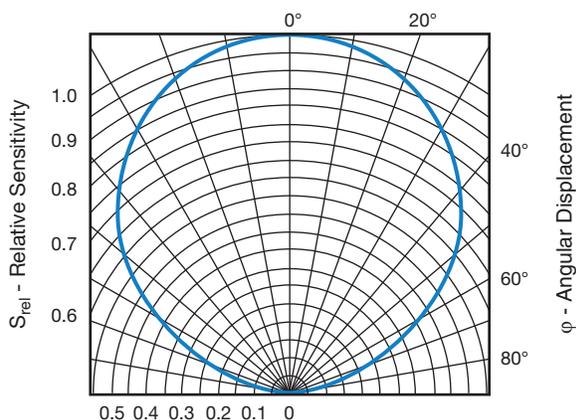


Fig. 6 - Angle of the Half Sensitivity of the Photodiode

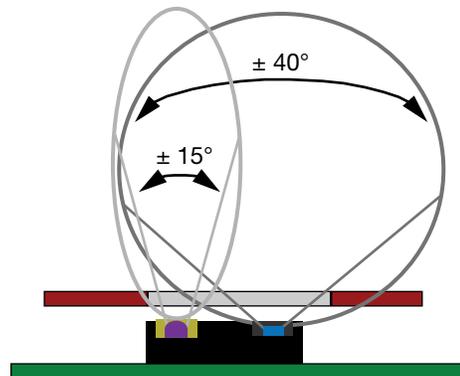


Fig. 7 - Emitter and Detector Angle

To achieve a good performance, the diameter of the hole within the cover glass should not be too small. An angle of  $\pm 40^\circ$  will be sufficient in most applications. The package drawing shows the position of the IRED and photosensitive area. The  $+40^\circ$  line should be set at the side of the photodiode, towards pin 1. The  $-40^\circ$  line should be set no closer than 1 mm to that edge. The following are dimensions for the distance from the top surface of the sensor to the outside surface of the glass,  $a$ , and the width of the window,  $d$ .

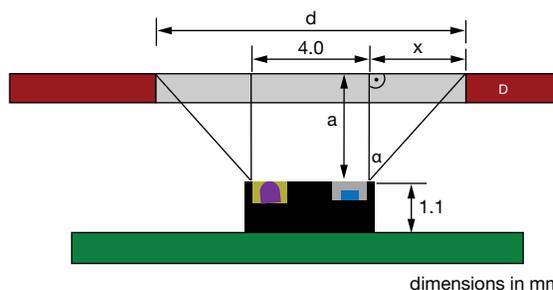


Fig. 8 - Window Dimensions for One Hole

For a single round hole, the diameter should be at least wide enough that the openings can freely look through; so, about 4 mm. Two much smaller holes are also possible. These should be at least the same as the diameter for the IRED = 1.2 mm.

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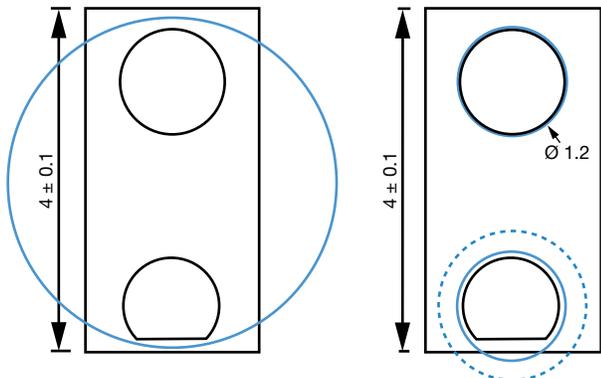


Fig. 9 - Light Holes Diameters

The diameter needs to be increased with distances between the sensor and cover glass according to the following calculation.

The width calculation for distances from 0 mm to 1.5 mm results with this in:

$$\begin{aligned}
 a = 0.0 \text{ mm} &\rightarrow x = 0.0 \rightarrow d = 4.0 \text{ mm} + 0.0 = 4.0 \text{ mm} \\
 a = 0.5 \text{ mm} &\rightarrow x = 0.42 \rightarrow d = 4.0 \text{ mm} + 0.84 = 4.84 \text{ mm} \\
 a = 1.0 \text{ mm} &\rightarrow x = 0.84 \rightarrow d = 4.0 \text{ mm} + 1.68 = 5.68 \text{ mm} \\
 a = 1.5 \text{ mm} &\rightarrow x = 1.28 \rightarrow d = 4.0 \text{ mm} + 2.56 = 6.56 \text{ mm}
 \end{aligned}$$

For the two smaller holes, the diameter for the IRED can be as small as 1.2 mm.

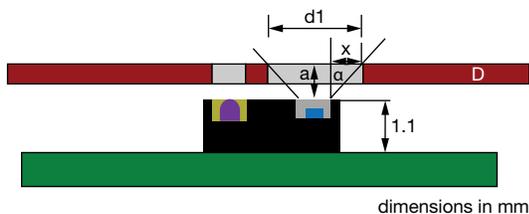


Fig. 10 - Window Diameters for Two Holes

Only the diameter for the photodiode needs to be increased, as shown in the example below, with distances between the sensor and cover glass.

The width calculation for distances from 0 mm to 1.5 mm results in:

$$\begin{aligned}
 a = 0.0 \text{ mm} &\rightarrow x = 0.0 \rightarrow d = 1.2 \text{ mm} + 0.0 = 1.2 \text{ mm} \\
 a = 0.5 \text{ mm} &\rightarrow x = 0.42 \rightarrow d = 1.2 \text{ mm} + 0.84 = 2.04 \text{ mm} \\
 a = 1.0 \text{ mm} &\rightarrow x = 0.84 \rightarrow d = 1.2 \text{ mm} + 1.68 = 2.88 \text{ mm} \\
 a = 1.5 \text{ mm} &\rightarrow x = 1.28 \rightarrow d = 1.2 \text{ mm} + 2.56 = 3.76 \text{ mm}
 \end{aligned}$$

The results above represent the ideal diameters of the window. The mechanical design of the device may not allow for these diameters.

### PROXIMITY SENSOR

The main DC light sources found in the environment are sunlight and tungsten (incandescent) bulbs. These kinds of disturbance sources will cause a DC current in the detector inside the sensor, which in turn will produce noise in the receiver circuit. The negative influence of this DC light can be reduced by optical filtering, but is reduced much more efficiently by a so-called DC kill function. The proximity photodiode shows its best sensitivity at around 850 nm, as shown in Fig. 11.

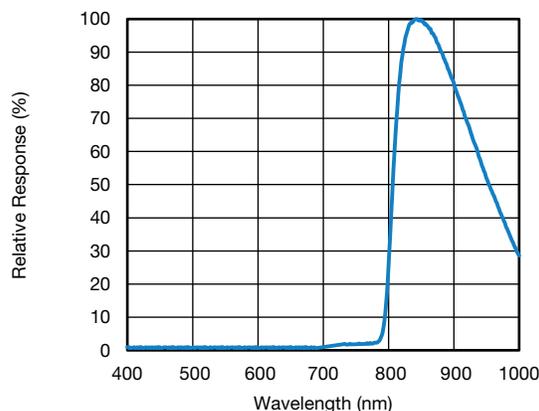


Fig. 11 - Spectral Sensitivity of the Proximity Photodiode

The proximity sensor uses a short pulse signal of about 125  $\mu\text{s}$  ( $\text{PS}_{IT} = 1\text{T}$ ) up to 1000  $\mu\text{s}$  ( $\text{PS}_{IT} = 8\text{T}$ ). The on / off duty ratio setting now defines which repetition rate to be used, which can be programmed from 1/40 up to 1/320.

In addition to DC light source noise, there is some reflection of the infrared emitted light off the surfaces of the components surrounding the VCNL3040. The distance to the cover, proximity of surrounding components, tolerances of the sensor, defined infrared emitter current, ambient temperature, and type of window material used all contribute to this reflection. The result of the reflection and DC noise is the production of an output current on the proximity and light sensing photodiode. This current is converted into a count called the offset count.

In addition to the offset count, there could also be a small noise floor during the proximity measurement, which comes from the DC light suppression circuitry. This noise is typically just one or two counts. Only with light sources with strong infrared content could it be in the range from  $\pm 5$  counts to  $\pm 10$  counts.

The application should “ignore” this offset and small noise floor by subtracting them from the total proximity readings. The VCNL3040 offers a subtraction feature what

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automatically does this: PS\_CANCEL. During the development of the end product, this offset count is evaluated and may now be written into register 5: PS\_CANCEL\_L/M. Now the proximity output data will just show the subtraction result of proximity counts - offset counts.

Results most often do not need to be averaged. If an object with very low reflectivity or at longer range needs to be detected, the sensor provides a register where the customer can define the number of consecutive measurements that the signal must exceed before producing an interrupt. This provides stable results without requiring averaging.

### PROXIMITY CURRENT CONSUMPTION

The VCNL3040 offers a shutdown mode. Default values after start-up has this disabled. The application needs to activate with PS\_SD = 0.

The VCNL3040's embedded LED driver drives the internal IRED with a pulsed duty cycle. The IRED on / off duty ratio is programmable by an I<sup>2</sup>C command at register PS\_Duty. Depending on this pulse / pause ratio, the overall proximity current consumption can be calculated. When higher measurement speed or faster response time is needed, PS\_Duty may be selected to a maximum value of 1/40, which means one measurement will be made every 4.85 ms, but this will then also lead to the highest current consumption:

PS\_Duty = 1/40: peak IRED current = 100 mA, averaged current consumption is 100 mA/40 = 2.5 mA.

For proximity measurements executed just every 40 ms: PS\_Duty = 1/320 peak IRED current = 100 mA, averaged current consumption is 100 mA/320 = 0.3125 mA.

The above is always valid for the normal pulse width of T = 1T = 125 μs, as well as for 2T, 4T, 8T, and all others in between. These pulse lengths are always doubled, resulting in 1000 μs for 8T, but the repetition time is also doubled, ending in a period time of about 320 ms.

An extremely power-efficient way to execute proximity measurements is to apply a PS active force mode (register: PS\_CONF3, command: PS\_AF = 1).

If only a single proximity measurement should be done, PS\_AF is set to "1" and then PS\_SD = 0 = active. Setting PS\_Trig = 1 will then execute just one single measurement.

In this mode, only the I<sup>2</sup>C interface is active. In most consumer electronic applications the sensor will spend the majority of time in sleep mode, it only needs to be woken up for a proximity or light measurement. In standby mode the power consumption is about 0.2 μA.

The pulse for proximity measurement looks to have a higher landing / step. This second trap is for smooth switch-off of the LED and is executed with very low IRED current. The pulse length in total is 200 μs. Amplitude of that first half is dependent on the IRED current. The higher this current is programmed, the higher that pulse amplitude will be. Taking a scope picture at IR\_Cathode (pin 4) will look like this:

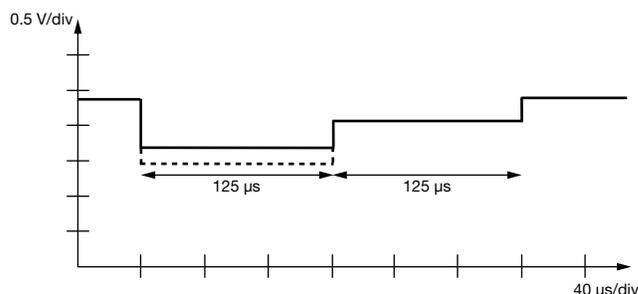


Fig. 12 - Proximity IRED Pulse for 1T

### INITIALIZATION AND I<sup>2</sup>C TIMINGS

The VCNL3040 contains seven 16-bit command codes for operation control, parameter setup, and result buffering. All registers are accessible via I<sup>2</sup>C communication. The built-in I<sup>2</sup>C interface is compatible with the standard and high-speed I<sup>2</sup>C modes. The I<sup>2</sup>C H-level voltage range is from 2.5 V to 3.6 V.

There are only five registers out of the thirteen that typically need to be defined:

1. LED\_I = 50 mA to 200 mA (IRED current)  
REGISTER PS\_MS #04 [0x04h]
2. PS\_Duty = 1/40 to 1/320 (proximity duty ratio), PS\_IT (proximity integration time = pulse length), PS\_PERS (number of consecutive measurements above / below threshold), and PS\_SD (PS power\_on)  
REGISTER PS\_CONF1 #03 [0x03h]
3. and 4. Definition of the threshold value from the number of counts the detection of an object should be signaled.  
Proximity TOP Threshold REGISTER PS\_THDL\_L #06 [0x06h] for the low byte and PS\_THDL\_H #07 [0x07h] for the high byte.

To define the infrared emitter current, as well as the integration time (length of the proximity pulsing), evaluation tests should be performed using the least reflective material at the maximum distance specified.

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Fig. 13 shows the typical digital counts output versus distance for three different emitter currents for integration time T1. The reflective reference medium is the Kodak Gray card. This card shows approximately 18 % reflectivity at 940 nm.

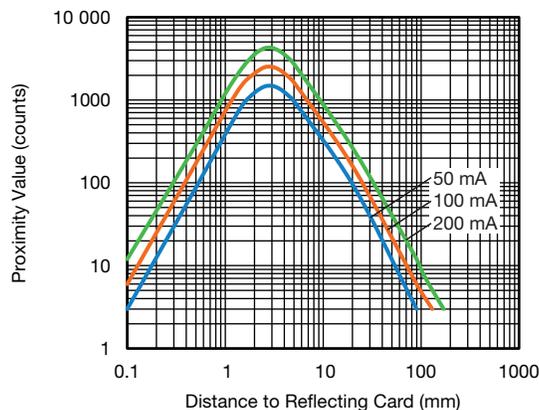


Fig. 13 - Proximity Value vs. Distance for 1T

This first diagram shows the possible detection counts with a short pulse of just 125  $\mu$ s.

If higher detection distances and / or objects with very low reflectivity should be detected, there is the option to extend these proximity pulses up to about 1000  $\mu$ s for 8T. This results in higher counts but may also lead to saturation effects for very close and very bright objects. This leads then to the diagram in Fig. 14 below.

In order to reach the high reflection counts of the Kodak Gray card, one has to define the proximity range to 16 bit, otherwise the 12-bit range would just lead to 4095 counts. This is possible to select with: PS\_HD = 1 within PS\_CONF3 byte of command code #3.

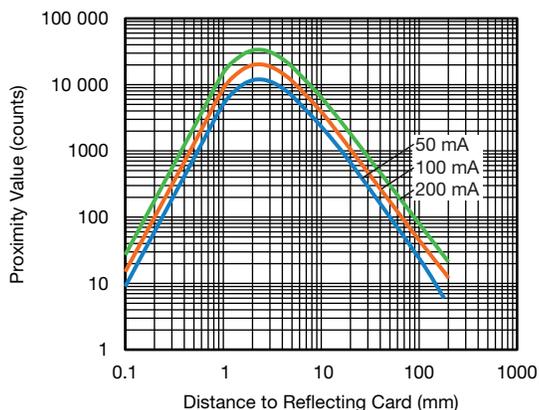


Fig. 14 - Proximity Value vs. Distance; PS\_IT = 8T

With defining the duty time (PS\_Duty), the repetition rate = the number of proximity measurements per second (speed of proximity measurements) is defined. This is possible between 5 ms (about 200 measurements/s) by programming PS\_Duty with 1/40 and 40 ms (about 25 measurements/s) with programming PS\_Duty with 1/320.

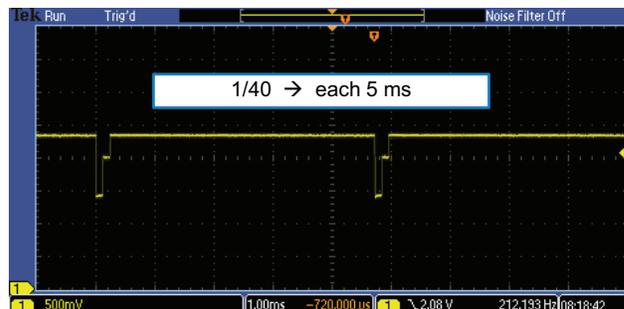


Fig. 15 - Proximity Measurements with PS\_Duty = 1/40

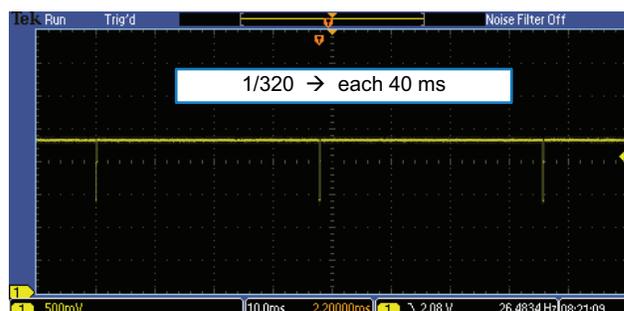


Fig. 16 - Proximity Measurements with PS\_Duty = 1/320

This duty cycle also determines how fast the application reacts when an object appears in, or is removed from, the proximity zone.

Reaction time is also determined by the number of counts that must be exceeded before an interrupt is set. This is possible to define with proximity persist: PS\_PERS. Possible values are from 1 to 4.

To define all these register values, an evaluation test should be performed. These tests can be made just using the VCNL4040 sensor board together with the SensorXplorer™. Both boards are available from any of Vishay's distributors.

Please see:

[www.vishay.com/optoelectronics/SensorXplorer](http://www.vishay.com/optoelectronics/SensorXplorer).

Please note that there is no dedicated VCNL3040 sensor board because all relevant settings and measurements are possible to realize with the VCNL4040-SB, just the "Ambient Light" measurements would lead to totally wrong numbers as this is not defined for the VCNL3040.



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### Timing

For an I<sup>2</sup>C bus operating at 100 kHz, to write or read an 8-bit byte, plus start (or stop) and bit acknowledgement, takes 100 μs. Together with the slave address byte and the 8-bit command code byte, plus the 16-bit data, this results in a total of 400 μs. When the device is powered on, the initialization with just these five registers needs 5 x 4 bytes (slave address, command register, and 16-bit data) for a total of 20 bytes. So, 20 x 100 μs = 2000 μs = 2 ms.

Send Byte → Write Command to VCNL3040



The read-out of 16-bit data would take a total of five bytes (slave address, command code, slave address with read bit set) and 16-bit data sent from the VCNL3040. So, 500 μs:

Receive Byte → Read Data from VCNL3040



### Power Up

The release of the internal reset, the start of the oscillator, and the signal processor need **2.5 ms**

### Initialize Registers

Write to four registers **1600 μs**

- IRED current
- Proximity duty ratio
- Proximity interrupt TOP threshold

Once the device is powered on and the VCNL3040 is initialized, a proximity measurement can be taken.

Asking for one forced proximity measurement **400 μs**

For (active forced, PS\_IT = 8)

Time to trigger [0.5 x PS\_IT] **500 μs**

DC-kill ambient light [3 x PS\_IT] **3000 μs**

Proximity measurement [1 x PS\_IT] **1000 μs**

IRED shutdown [1 x PS\_IT] **1000 μs**

Read out of the proximity data **500 μs**

total: **6400 μs**

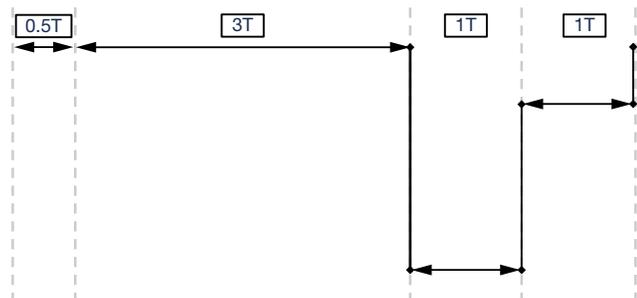


Fig. 17 - Timing Specification for Active Forced Mode

### Interrupt

The VCNL3040 features a very intelligent interrupt function. The interrupt function enables the sensor to work independently until a predefined proximity event or threshold occurs. It then sets an interrupt which requires the microcontroller to awaken. This helps customers reduce their software effort, and reduces power consumption by eliminating polling communication traffic between the sensor and microcontroller.

The interrupt pin, pin 7, of the VCNL3040 should be connected to a dedicated GPIO of the controller. A pull-up resistor is added to the same power supply that the controller is connected to. This INT pull-up resistor may be in the range of 8.2 kΩ to 100 kΩ.

A lower and an upper threshold for the proximity value can be defined. If the proximity value falls below the lower limit or exceeds the upper limit, an interrupt event will be generated. In this case, an interrupt flag bit in the read-out register 0x0B will be set and the interrupt pad of the VCNL will be pulled to low by an open drain pull-down circuit. In order to eliminate false triggering of the interrupt by noise or disturbances, it is possible to define the number of consecutive measurements that have to occur before the interrupt is triggered.

Beside this “normal” interrupt mode, an automatic mode is also available, which is called the logic output mode.

This mode automatically pulls the interrupt pin low when an object exceeds the programmed upper threshold and also resets it if the lower threshold is exceeded. So no actions from the controller are needed if, for example, a smartphone is held close to an ear but quickly taken away (e.g. for a short look at the display).

### Application Example

The following example will demonstrate the ease of using the VCNL3040 sensor.

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### Offset

During development, the application-specific offset counts for the sensor were determined. As previously mentioned, the offset count is affected by the components surrounding the VCNL3040, the window or cover being used, the distance from the sensor to the cover, and emitter intensity, which is controlled by the forward current.

In the following example, with a cover over the sensor and setting the emitter current to 50 mA, the offset counts are 540 counts (Fig. 21). Offset counts vary by application and can be anywhere from 0 counts to several thousand counts. It is important to note that the offset count may change slightly over time due to, for example, the window becoming scratched or dirty, or being exposed to high-temperature changes. If possible, the offset value should occasionally be checked and, if necessary, modified.

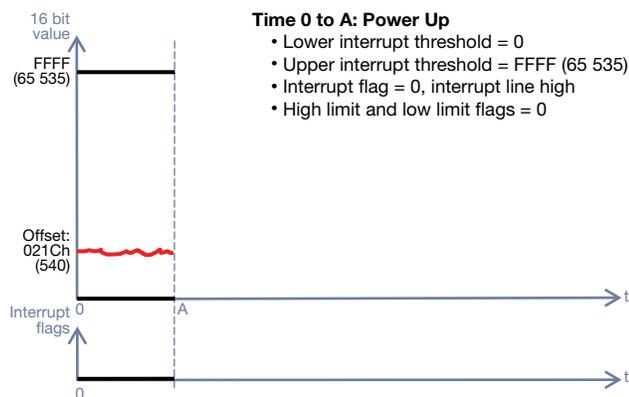


Fig. 18

### Power Up

As mentioned, there are four variables for proximity measurement that need to be set in the register when the sensor is powered up: the emitter current, the number of occurrences that must exceed a threshold to generate an interrupt, the threshold values, and the number of proximity measurements per second.

The sensor should detect skin at a distance of 5 cm. Development testing determined that a current of 50 mA produces adequate counts for detection. The proximity measurement rate is set so that about 10 measurements are done within a second and the number of occurrences to trigger an interrupt is set to four. Based on development testing, with a hand or skin approximately 5 cm above the window cover, the resulting total count is > 550. This will be used as the upper threshold (high threshold).

For smartphone applications it would be typical to initially set this top threshold and a lower threshold (bottom threshold). This is needed to indicate the removal of the phone from the user's ear. The measured counts without any additional object close by will be around this offset count value, always below the lower threshold value, as shown in Fig. 22.

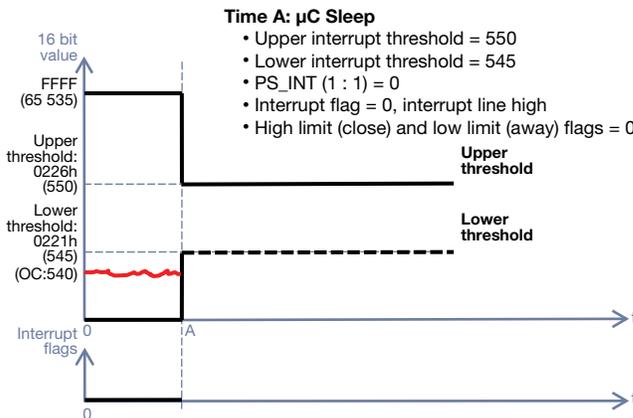


Fig. 19

By setting the number of occurrences before generating an interrupt to 4, a single proximity value above or below the thresholds will have no effect, as shown in Fig. 23.

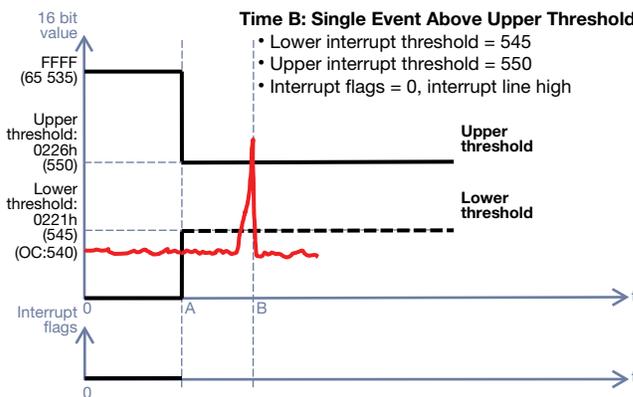


Fig. 20

A smartphone application will use a proximity sensor to detect when the phone is brought to the user's ear and disable the touchscreen and turn off the backlight. For other applications, such as automatic dispensing, the soap or towel will be dispensed.



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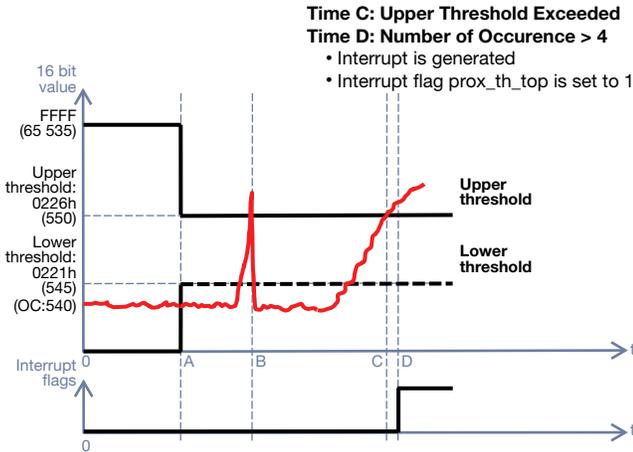


Fig. 21

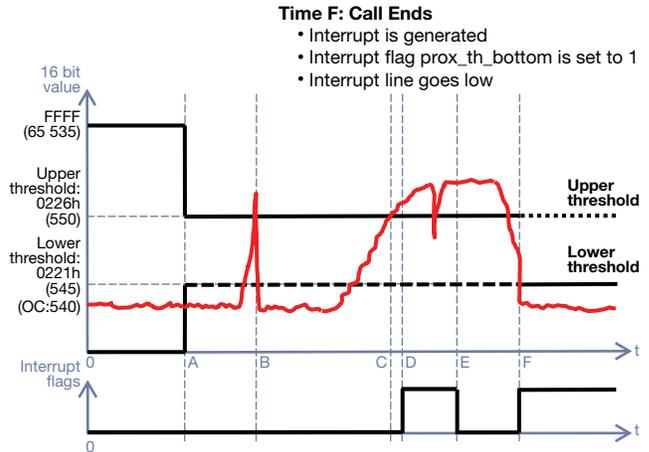


Fig. 22

In smartphone applications, the bottom threshold will also be programmed and waits for an interrupt signal. The prox\_threshold\_bottom should be set to “1” now and the prox\_threshold\_top cleared by entering a “1” again, since the phone is already next to the user’s ear. A lower threshold will occur when the phone call is complete and the phone is brought away from the user’s ear, and the backlight and touchscreen will be turned back on.

For this example, the upper threshold will only be set to 560 counts. The lower threshold is set to 545 counts; a value that is higher than the offset but low enough to indicate the removal of the phone from the user’s ear.

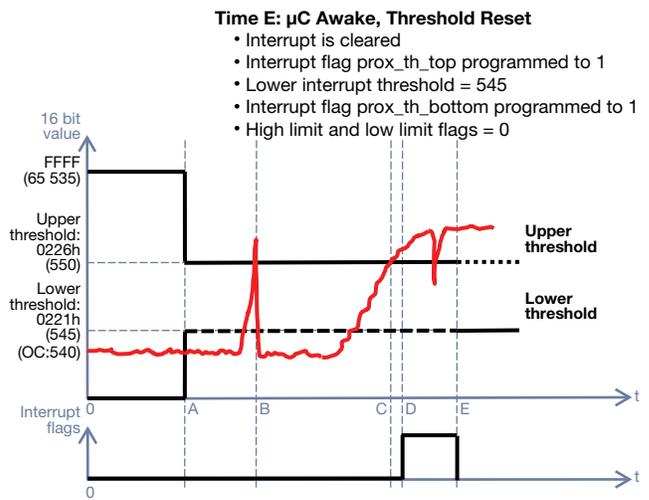


Fig. 23

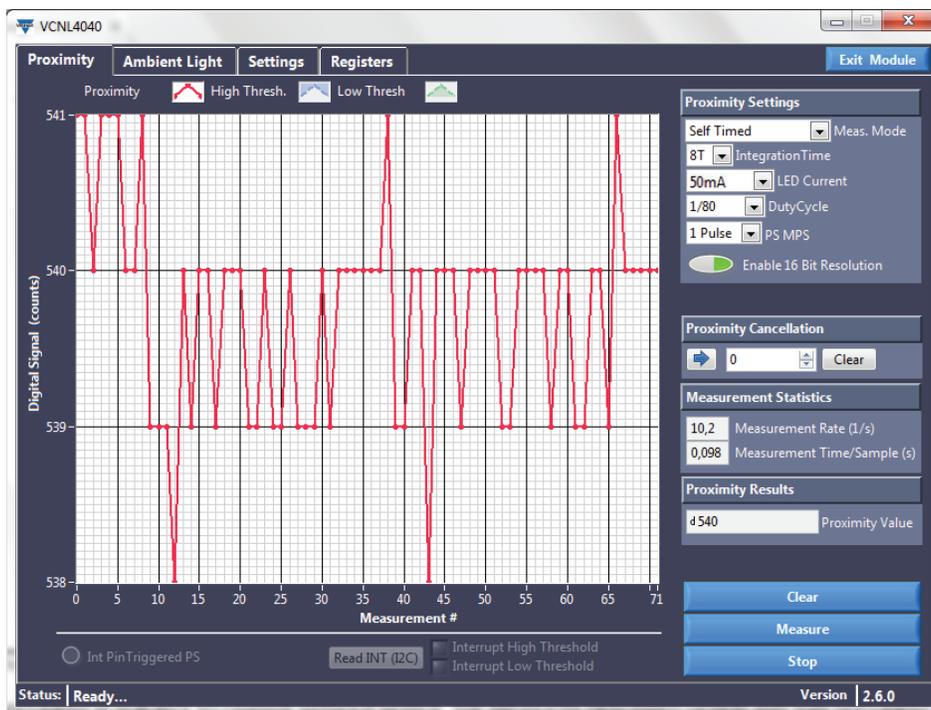


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## Designing the VCNL3040 Into an Application

Some measurements and features are shown with the demo tool and demo software with a cover glass at about a 2 mm distance.

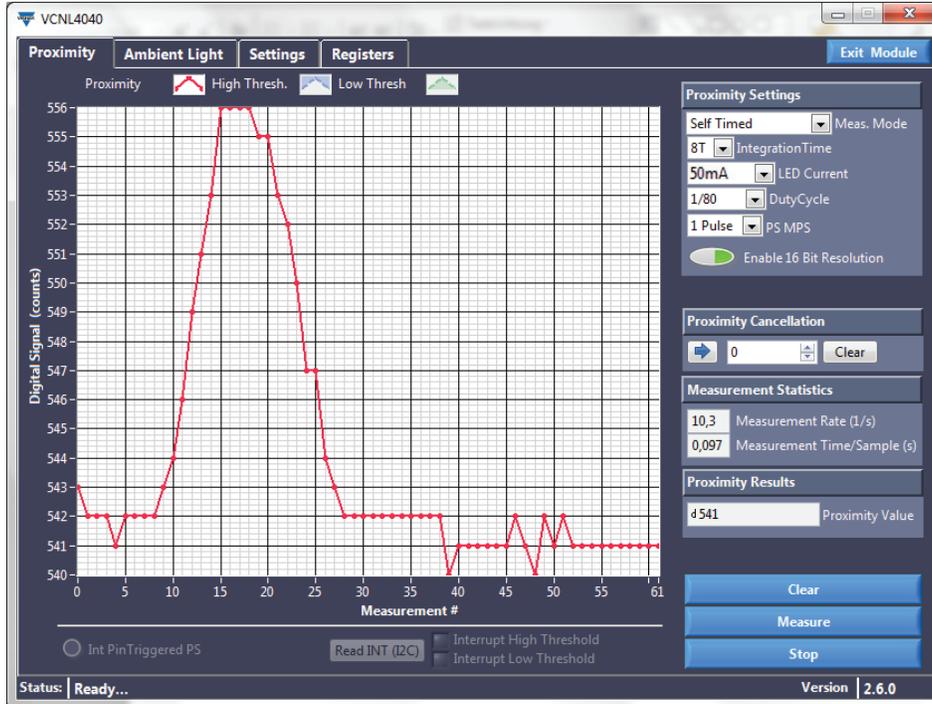
1. Proximity set-up with 8T wide pulses, 50 mA emitter current, and a duty cycle of 1/80, which results in about 10 measurements per second



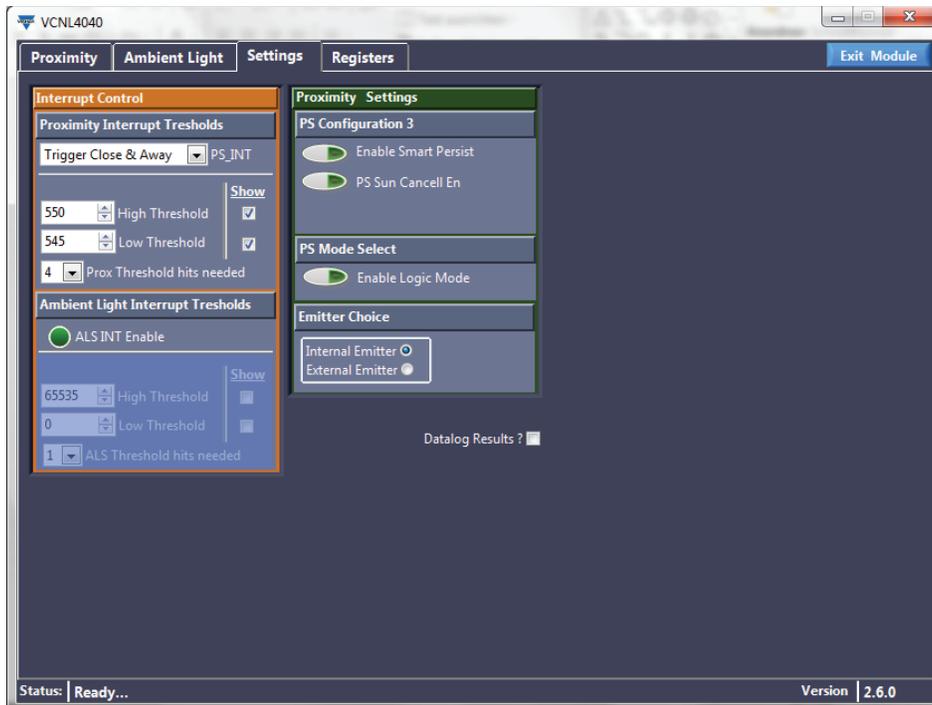
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2. If a hand or skin now comes as close as 5 cm, these 540 counts rise up to more than 550 counts.



3. Here the thresholds are programmed as 550 for the upper and 545 for the lower. To see these, both “Show” buttons are activated. The presence of an object should only be recognized when four consecutive measurements are above that threshold.

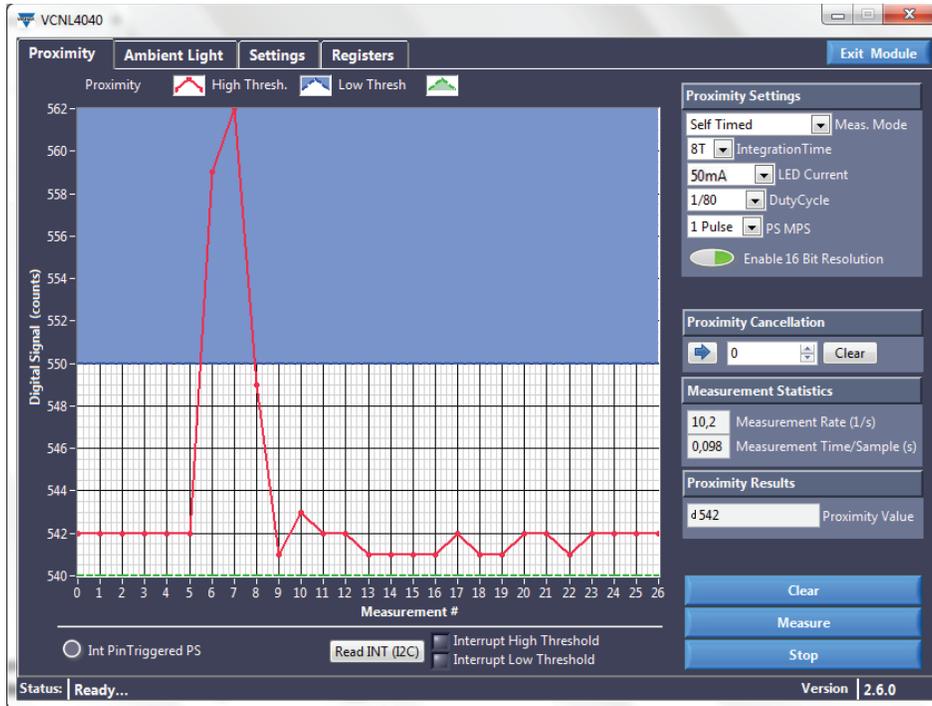




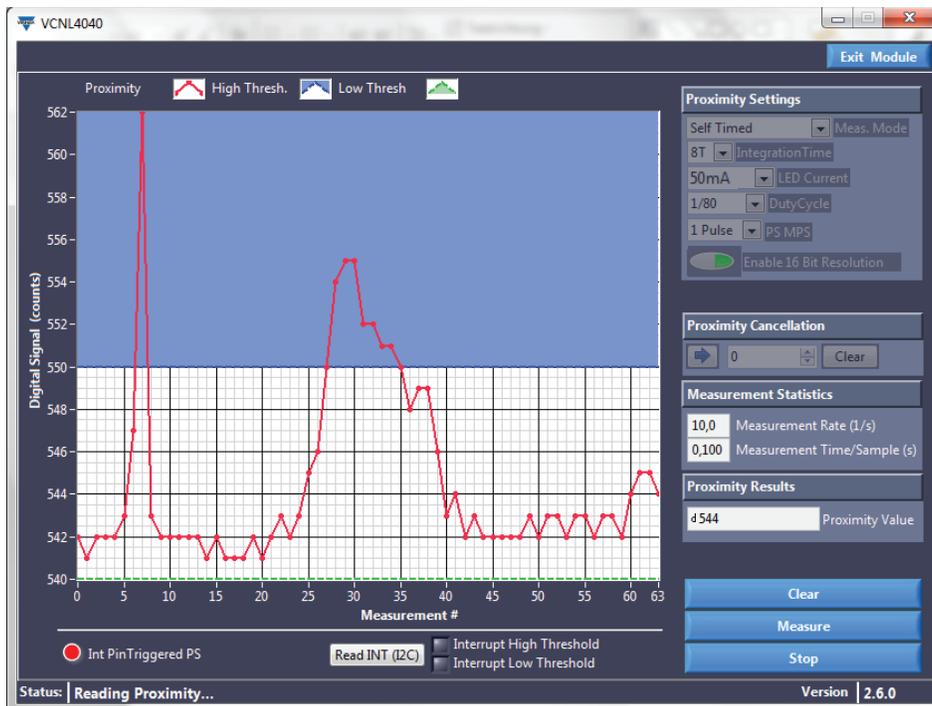
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4. Just one or two measurements above the threshold will not activate the interrupt.



5. With more than four measurements above the threshold, however, the interrupt is pulled low, as indicated by the red LED on the demo board and the red light: “Int Pin Triggered PS.”

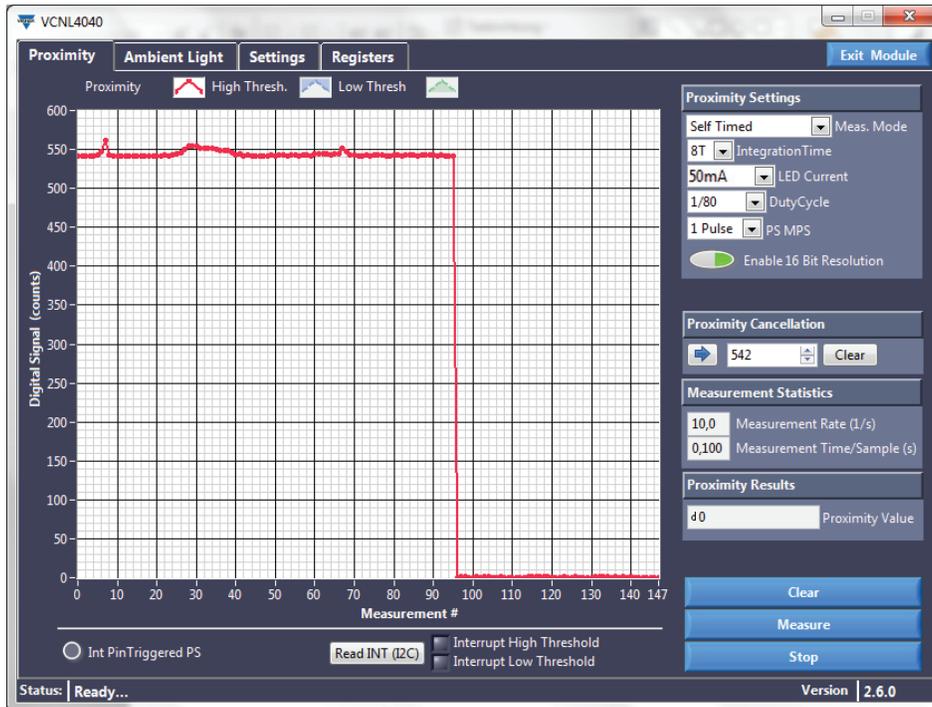


APPLICATION NOTE

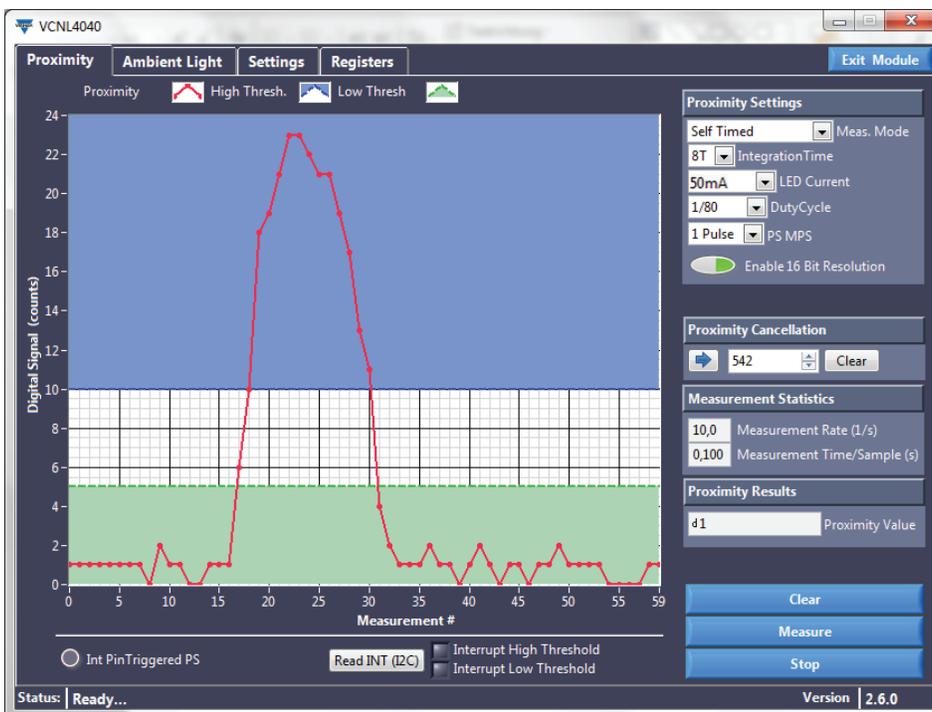
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6. The cancellation feature is used below. The “before seen” offset counts are subtracted. To do so, the value of 540 is entered for register number 05 = Prox\_Cancellation.



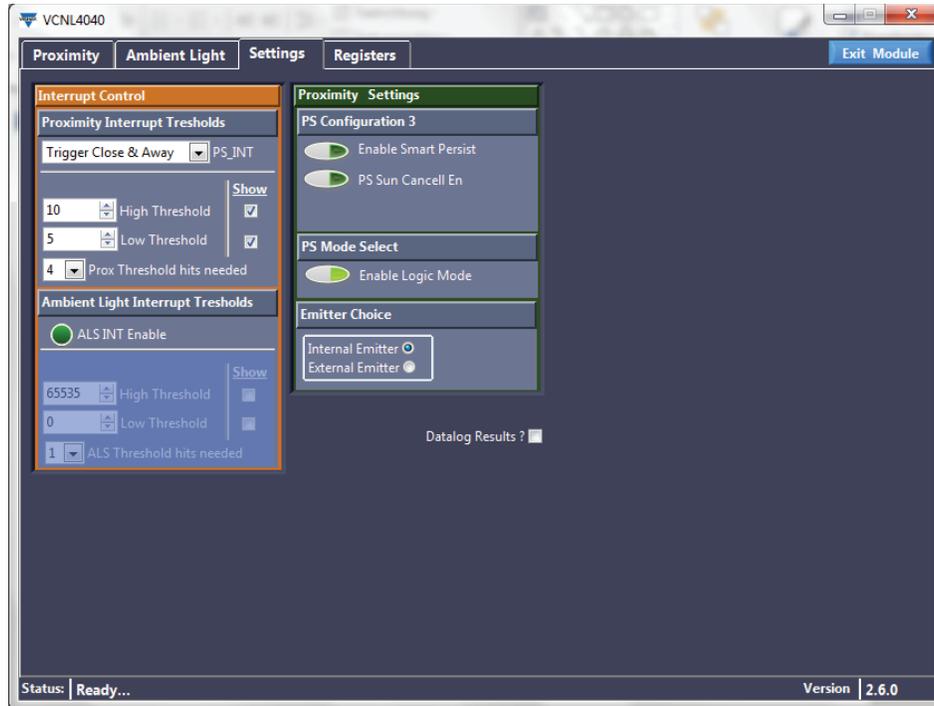
7. The “before seen” measured proximity result data of 540 is now  $540 - 540 = 0$ . Also, the thresholds are now 540 counts lower. The higher threshold is 10 and lower is just 5.



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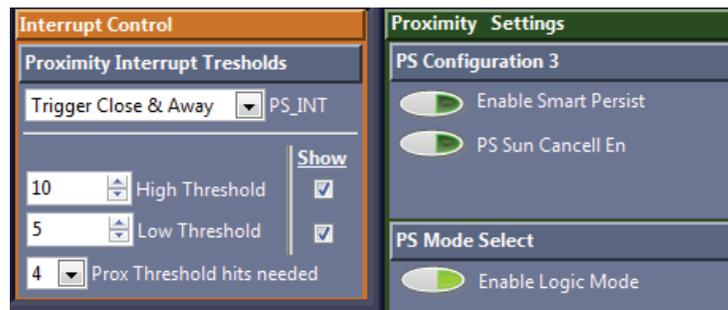
If one chooses “logic mode” now and redefines the high threshold to 10 and low threshold as 5...



... the interrupt will indicate the rise above the upper threshold and will also automatically be cleared when it falls below the lower threshold.

One special feature for faster proximity measurements is also implemented, which is called “smart persist.”

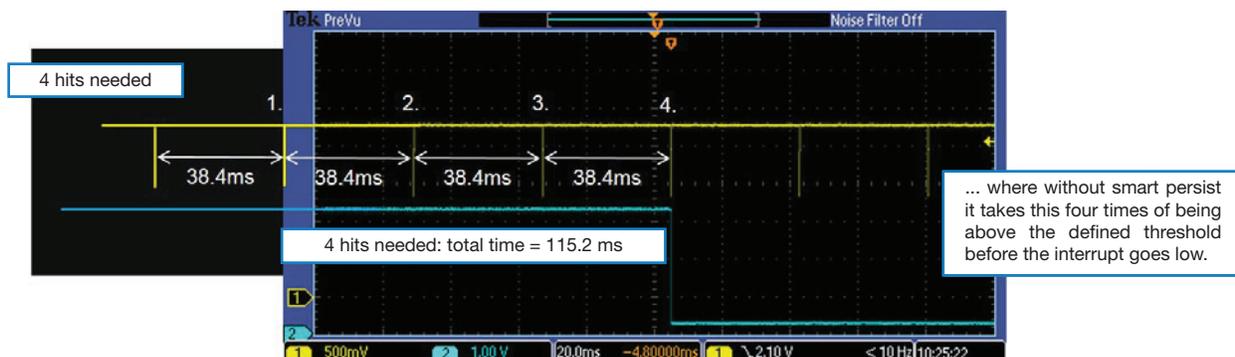
This feature reduces the total reaction time until the interrupt is set to active, although four consecutive measurements should be above (or below) the defined threshold for safe acknowledgment.



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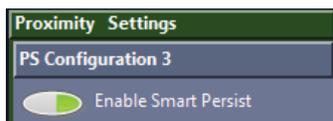
Without “smart persist”, but with programmed hits above the defined threshold set to four, it will take four times the time of PS\_Duty. With PS\_Duty set to 1/320 this would be 4 x 38.4 ms.



With “smart persist” activated (bit 4 of PS\_CONF3):

REGISTER: PS_CONF3 DESCRIPTION		
REGISTER: PS_CONF3		COMMAND CODE: 0x04_L (0x04 DATA BYTE LOW)
Command	Bit	Description
Reserved	7	0
Reserved	6 : 5	(0 : 0)
PS_SMART_PERS	4	0 = disable; 1 = enable PS smart persistence

or within the demo-tool:



The total needed time is reduced to just one time of 38.5 ms, followed by three times of just 1.3 ms between the next three measurements, for a total of 39.7 ms.



**Remark:**

With “smart persist” enabled, there will always be four pulses shortly after each other, whether PS\_PERS is set to 2, 3, or 4.